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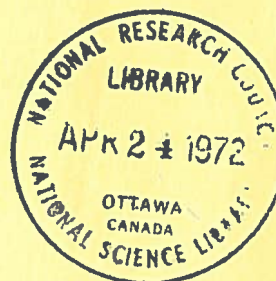
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NATIONAL RESEARCH COUNCIL OF CANADA

RADIO BRANCH

THREE CENTIMETRE EXPERIMENTAL SET  
PRELIMINARY REPORT



OTTAWA

FEBRUARY, 1944

This report has reference to the package sets produced at the National Research Council of Canada, Radio Branch, and is not intended to describe the production sets known as RX/F.

THREE CENTIMETRE EXPERIMENTAL SET  
PRELIMINARY REPORT

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NATIONAL RESEARCH COUNCIL OF CANADA  
RADIO BRANCH

THREE CENTIMETRE EXPERIMENTAL SET

PRELIMINARY REPORT

I - INTRODUCTION

(1) Purpose

This is a microwave search set, operating in the three-centimetre band. It is designed primarily for use with Motor Torpedo Boats, but can readily be fitted to most types of small surface craft. Its function is the detection of other surface craft, including surfaced submarines; and it is believed that it will prove useful for navigation in coastal waters.

(2) Performance

The set is designed to give range accurate to  $\pm 200$  yards, and bearing accurate to  $\pm 2^\circ$ . Range scales of 7500 and 30,000 yards are available at the option of the operator on both a Plan Position Indicator, and a Class A display. The PPI unit is fitted with a cursor and magflip transmitter so that bearing may be set and automatically relayed to any remote point. On the basis of trials made with a laboratory model, ranges of the following orders may be expected when the aerial is fifteen feet or so above water line:

Surfaced submarines	-	5,000	-	7,000 yards
M.T.B.'s	-	5,000	-	7,000 yards
Fairmiles	-	8,000	-	11,000 yards
Corvettes	-	11,000	-	14,000 yards
Medium Merchantmen	-	15,000	-	20,000 yards

(3) Layout

The set has six components of size and weight approximately as shown below:

<u>UNIT</u>	<u>APPROX. WEIGHT</u>	<u>APPROX. DIMENSIONS</u>
Antenna-Rotator	100 lbs.	(Antenna: 30" x 20" x 12" high (Rotator: 14" x 11" x 12" high
R.F. and Modulator	100 lbs.	14" x 14" x 20" high
Monitor	Monitor 45 lbs. (Power Supply 50 lbs.)	19" x 15" x 11" high 15" x 12" x 11" high



<u>UNIT</u>	<u>APPROX. WEIGHT</u>	<u>APPROX. DIMENSIONS</u>
Plan Position Indicator	50 lbs.	15" x 12" x 11" high
Control Box	10 lbs.	11" x 10" x 5" high

Antenna-Rotator - The antenna is a 30-inch parabolic cylinder, fed by an electromagnetic horn. It is mounted on a motor-driven shaft rotating at approximately 22 r.p.m. The motor, reduction gear and transmission line couplers are enclosed in a water-tight cast aluminum housing. This housing also contains a Selsyn generator which controls the rotating sweep of the PPI. The housing is equipped with holes for mounting-studs, and may be attached to a tripod or other supporting device appropriate to the vessel on which the set is to be installed. Ref. dwg. RX/F-49-E and dwg. RX/F-375-D.

R.F. and Modulator Unit - This unit consists of three chassis, r.f., Modulator and Automatic Frequency Control, mounted on a single rack in a splash-proof case. The A.F.C. unit consists of a discriminator, controlling the frequency of the beat oscillator, and contains also the power supplies for the beat oscillator and pre-amplifier. Ref. dwgs. RX/F-176-B, RX/F-173-B and RX/F-171-B.

Monitor - This unit contains the main receiver and sweep circuits. The latter generates sweep voltages for both a 5-inch type A display tube, and a 5-inch PPI tube. The linear display tube is in the monitor chassis and is intended for general tuning-up and observation of performance of the set. The range scales are 7500 and 30,000 yards, with 1000 and 5000 yard calibration pips. Ref. dwgs. RX/F-166-B and RX/F-164-A.

Monitor Power Supply - The power supply for the monitor cathode ray tube, PPI cathode ray tube, and scanning circuits is included in a separate chassis. Ref. dwg. RX/F-163-A.

Plan Position Indicator (PPI) - The PPI is a 5-inch long delay tube enclosed in a splash-proof case. It has range scales of 7500 and 30,000 yards, with 1000-yard and 5000-yard calibration circles. The latter may be turned on or off at the operator's option. Azimuth is determined from a rotatable cursor with a 180°-0-180° engraved scale. It is equipped with a Selsyn, Magslip transmitter or "M" type switch, by means of which azimuth information may be relayed to any remote point. Ref. dwgs. RX/F-167-B, RX/F-124-E and RX/F-230-D.

Control Box - This is a splash-proof unit containing a low-voltage a.c. and d.c. switch, high-voltage switch, and a Variac controlling the pulse voltage. Ref. dwg. RX/F-324-A.

Location of Units - The antenna-rotator may be mounted in any appropriate location by means of studs. On Motor Torpedo Boats it is anticipated that this unit will be mounted on a steel tripod above the cabin at an approximate height of fifteen feet above the water-line.



The r.f., Modulator, Monitor and Control Box unit can be installed below deck in any convenient location. On M.T.B.'s it is expected that this will be in the W/T room. The R.F. unit is connected to the antenna by a rectangular waveguide transmission line. To keep transmission-line losses to a reasonable figure, this distance should not exceed fifteen feet or so, and where possible should be less. Two flexible cables carrying Selsyn leads run between the Antenna-Rotator and the Monitor Unit. One of these cables also carries leads for the d.c. power to the driving motor.

The PPI unit may be located wherever desired. It is assumed that this will usually be in the wheelsman's or navigator's cabin. A flexible cable connects the PPI with the Monitor.

Operators - Two operators are required. One is a technical operator to handle the Control Box, R.F., Modulator and Monitor units. The second is an observer for the remote PPI.

Power Requirements - The total requirements of the set are 1.5 kilowatts at 115 volts, 500 cycles, and 400 watts at 110 volts d.c.; the latter being for the antenna drive, motor blowers, and spark wheel motor. Both types of power are fed to the Control Box which is the distributing centre of the set.

## II - SEQUENCE OF OPERATION

(see Block Diagram RX/F-175-B)

The modulator consists of a rotary spark wheel which discharges at roughly 500 times a second, a resonantly charged line. The pulse is about 3/4 of a microsecond duration, 15 kv. at 12 amperes. This is delivered to a 2J21 magnetron which is waveguide coupled to a parabolic slice antenna through the antenna gear box. A single antenna for transmitting and receiving is used. Protection is provided by means of an RT/TR system which has the advantage over a single TR in that it is independent of the cold impedance of the magnetron. The local oscillator is a W.E. 723-A tuned by varying its reflector bias. Automatic frequency control is provided in such a way that varying frequencies through the I.F. pre-amplifier pass through a discriminator circuit of conventional design, followed by a d.c. amplifier. Depending on the direction of the frequency shift, proper compensating voltages are applied to the oscillator reflector.

The pulse to the magnetron from the modulator also triggers a multivibrator which provides a square wave pulse which is placed on the grid of a power tube acting as a cathode follower. In the cathode circuit are connected the rotor coils of a 3" Selsyn which is placed in the antenna gear box. The stator of this same Selsyn is coupled to a similar stator placed around the neck of a 5" cathode ray tube (long persistence). This is the "remote PPI" unit. The square voltage wave is thus effectively placed on the Selsyn



inductance, providing a saw tooth of current which forms the sweep. As the antenna rotates, the magnetic field also rotates and moves the sweep around the tube face. Centering is provided by a negative current pulse through the Selsyn immediately after the sweep. The current sweep through a small resistance is voltage amplified to provide a saw tooth voltage sweep for monitoring purposes. The same pulse that begins the sweep shocks a tuned circuit, which provides pips for calibration purposes. There are two sweep lengths: 7500 yards and 30,000 yards.

The receiver is a wide-band (10 mc.) I.F. followed by a narrow band (.5 mc.) video.

As shown in the block diagram there are six units, each splash-proof and cable connected.

### III - ANTENNA

#### (1) Theory (see RX/F-168-A)

The rectangular waveguide (1" x 1/2" O.D.) which emerges from the  $E_{01} - H_{01}$  transformer in the gear box is twisted through 90° and then bent in a semi-circle of 4" radius. After the bend the waveguide is terminated in a flared horn. The horn proper is 3-3/8" in depth, with inside throat dimensions of 7/8" x 3/8", and inside mouth dimensions of 4-3/8" x 0.197". The overall width of the mouth is 9/16" x 6-1/4" due to flanges on each side and end of the horn. A mica window, 0.010" ± .001 thick is clamped to the front of the horn by a brass window frame 1/8" thick which is screwed down firmly onto the mica. The edges then are soldered (so the solder runs into the edge of the mica) to prevent leakage of radiation.

The mirror is a section of a 38.8" parabolic cylinder, cut off in the plane where the diameter would be 36", and bent in at the ends to give an aperture diameter of 30". This means that the focus is outside the aperture plane and  $f = .315 D$ , where  $D$  is the diameter of the aperture. The height of the section is 6-1/4".

The front edge of the window frame on the horn - that is, the extreme front section of the horn - should be 9.640" ± 0.020 from the vertex of the mirror.

#### (2) Rotational Equipment (see RX/F-169-B)

(a) Electrical - A waveguide line throughout is used between magnetron and antenna. The greater part of this consists of rectangular waveguide 1" x 1/2" O.D. with 0.064" walls. At 3.20 cm. only the  $H_{01}$  mode is propagated, the guide wavelength being 4.66 cm. A rotating joint in waveguide is required, and for this section of the line circular guide transmitting the centrally



symmetrical  $E_{01}$  mode is used. This requires two "wave type transformers" to transform from the  $H_{01}$  mode in the rectangular guide to the  $E_{01}$  in the circular, and back again to  $H_{01}$  in the rectangular for the short run to the antenna. The type of transformer adopted for this set is shown in Drawing RX/F-169-B.

For the circular waveguide brass tubing 1-1/8" I.D. (1-1/4" O.D.) is used. At 3.20 cm. such guide transmits a  $H_{11}$  mode (guide wavelength =  $\lambda_g = 4.24$  cm.) and an  $E_{01}$  mode ( $\lambda_g = 6.21$  cm.).

In the present design of rotating coupling the  $H_{11}$  mode is eliminated by making the stub end of the circular guide the proper length. If the distance from the end of the circular guide to the mid-line of the rectangular guide is  $3/4 \lambda_g$  for the  $H_{11}$  mode, a high impedance will be presented to the  $H_{11}$  mode and this will not be set up appreciably unless the overall length of the rotating assembly is resonant to the  $H_{11}$  mode. To avoid this the overall length should be an odd number of quarter guide wavelengths for the  $H_{11}$  mode. Through an oversight the overall length of the present unit was made  $3/8$ " shorter than had been intended. A careful test of this actual length showed no trace of  $H_{11}$  resonance, which in this case would be expected to occur near 3.20 cm. This presumably is because the stub ends are adjusted to the optimum length for eliminating the  $H_{11}$  mode at this wavelength. Also the band pass was not seriously worse than for the intended length so it has been left unaltered. The dimensions shown on RX/F-169-B are the dimensions the present unit actually has.

When the length of the stub end is  $3/4 \lambda_g$  for the  $H_{11}$  mode, it will be at the same time approximately  $1/2 \lambda_g$  for the  $E_{01}$  mode in 1-1/8" circular guide. This makes the entrance to the circular guide as viewed from the rectangular a low impedance point for the  $E_{01}$  mode. The residual reflection at the junction gives a S.W.R. of 1.38 which is matched out by inductive irises in the rectangular guide. The following tables give measured standing wave ratios for an experimental model of the present complete unit.

$\lambda$	S.W.R. (volts)
3.17	1.16
3.18	1.09
3.19	1.07
3.20	1.02
3.21	1.02
3.22	1.07
3.23	1.17

The ends of the circular guide at the joint are fitted with chokes to prevent radiation from the gap. Measurements have shown such chokes transmit 97% of the power when the gap is 1 mm. A similar choke is used for mechanical reasons at the bottom of the rotator. In each case the gap is approximately 1/32"



Measurements on a model of the complete unit have shown that, with no gaps, 95% of the power is transmitted from rectangular guide to rectangular guide, or a loss of 0.2 db.

(b) Mechanical (RX/F-49-E)

The gear box proper is totally enclosed and water-tight and is powered with a 1/10 h.p., 3500 r.p.m., d.c. motor. Output speed is approximately 22 r.p.m.

Internal Components:

- (i) Motor Shaft - The motor is flange mounted in the main casting and carries a 12 P. double shaft hardened and polished steel worm, keyed to its shaft. This worm meshes with a bronze 12 P. worm wheel on the countershaft.
- (ii) Countershaft - The countershaft mounts in addition, an 8 P. hardened and polished steel worm, one single row unrestrained ball bearing and a double row pre-loaded bearing to restrain the end thrush on the shaft.
- (iii) Main Shaft - The main shaft carries a bronze 8 P. worm wheel and is mounted on two tapered roller bearings. Cut integral with the shaft is an 8 mm. pitch sprocket to facilitate chain drive to the Selsyn generator. Keyed and screwed to the upper end of the shaft is the antenna mounting flange.
- (iv) Selsyn Mounting - The Selsyn mounting is of the flanged type, this flange being eccentric to the centre line of the Selsyn to allow proper adjustment of the driving chain. The cup shaped sprocket is pressed into a double row shielded ball bearing, which in turn is rigidly held in the mounting proper. A keyed sleeve mounted on the Selsyn shaft mates with keyways cut in the sprocket shaft, thus enabling the Selsyn to be removed without disturbing the driving mechanism.

Water-proofing:

All removable covers are sealed with "Velumoid" gaskets. The upper end of the main shaft is sealed with a "Garlock Klosure" seal. In addition the labyrinth between the main housing and the skirt of the antenna mounting flange is filled with grease. The power and Selsyn leads are taken through glands on either side of the housing.

Mounting:

The rotator may be mounted either by the four base mounting holes or by the four back mounting holes or both sets of holes. In both cases studs are to be screwed into the main casting in the holes provided. In no event should cap screws be used, as the aluminum threads will become worn through tightening and loosening of the screws.



Lubrication:

To fill the gearbox to the required level two imperial quarts of oil are required. For temperature ranges below 32°F one of the following oils should be used: DND 360-A (87-1/2%) and Kerosene (12-1/2%), 3-G.P.-26 or SAE 10W (80%) and Kerosene (20%). Above 32°F any one of the following oils may be used: DND 360-A or SAE 10W.

Note - Care should be taken to avoid attempting to mix DND 360-A with other lubricating oils than Kerosene.

IV - R.F. SYSTEM

(1) General

(a) Introduction - (see RX/F-172-A)

The magnetron transmitter is a Raytheon 2J21 which operates at about 2500 oersteds magnetic field, 10 - 14 kv., peak plate current of 10-12 amps., and output 20 kw. peak. It is supplied with a waveguide output fitting which feeds directly to the waveguide in the set through a flange and choke coupling. For the best performance of the magnetron the magnetic field must be in the correct direction. This is indicated on each data sheet (i.e. cathode south) and provision is made so that the magnet may be reversed by loosening two thumb screws which release the magnet from its mount.

The R.F. power is conducted through the TR/RT system to the gear box in rectangular brass waveguide 1" x 1/2" O.D. x .064" wall thickness. Water-tight choke couplings are used to connect various pieces of the waveguide. Fixed matching in the case of the horn, the  $H_{01} - E_{01}$  transformer, and the TR and RT is accomplished by means of inductive diaphragms.

(b) Duplexing System - (see RX/F-174-A)

The duplexing system employs both a TR and an anti-TR (or RT) switch. These switches are similar in construction except that the RT switch has no output window, and its cavity and input window are slightly larger than in the TR switch. Both switches use a Sylvania 724-A filler in a cylindrical cavity, and are window-coupled to the waveguide. They are tuned by two adjustable plungers on diametrically opposite sides of the cavity. These plungers are fitted with chokes to prevent losses, and give a tuning range of approximately 1%. A negative potential of about 600 volts on the "keep-alive" electrode of the TR switch gives, through a suitable series resistance, a "keep-alive" current of approximately 0.1 ma. No "keep-alive" potential is necessary with the RT switch.

An E-plane, series branching system is employed. The input windows of the TR and RT are approximately  $\lambda_g/2$  from the inner wall of the main guide, and the centres of the TR and RT branches are about  $\lambda_g/4$  apart.

On transmission the mismatch caused by the duplexing system gives a standing wave ratio of about 1.2 (power) and the leakage power to the crystals of the order of 70-100 mw. (peak).



On reception the total loss through the duplexing system is slightly less than 2 db., while the mismatch causes a standing wave ratio of the order of 1.75 (power).

(2) Local Oscillator

The local oscillator is a Western Electric 723-A oscillator which operates at a resonator voltage of 300 volts, filament voltage 6.3 volts, and reflector voltage from -20 to -300. In the present set the reflector voltage is usually between 100 and 200 volts. Local oscillator power to the crystal is controlled by the screw adjustment of the projection of the output spike into the waveguide. It has been found that a crystal current of from 300 to 500 microamps. is best for the average crystal. The 723-A may be tuned either mechanically (range: 3.16 cm. to 3.24 cm.) or electrically, by changing the reflector voltage. ( $\Delta f$  is 2 to 5 mc. per second per volt). In the present set the mechanical tuner is first set so that the desired frequency can be reached with electrical tuning. Tuning in operation is then accomplished by an automatic frequency control; or by throwing a switch, tuning may be accomplished electrically by means of a potentiometer. (See A.F.C. below).

(3) Mixer (see RX/F-170-B)

The mixer consists of a standard B.T.H. crystal held in a special holder and plugged in to extend across the waveguide. Matching to the waveguide is done with a screw which can be adjusted as to length and position. The I.F. output from the crystal is soldered directly to the input of the pre-amplifier.

(4) Pre-amplifier (see RX/F-171-B)

The pre-amplifier input tuned circuit is loaded only by the crystal and first I.F. amplifier tube grid, in order to get the best signal-to-noise ratio possible. Three stages of amplification follow, employing resistor loading and unity coupled interstage transformers. Tuning is accomplished by a brass slug screwed into the centre of each interstage transformer, and into the input and output coil.

The third stage has a low impedance output to drive the concentric line feeding the remainder of the receiver in other boxes.

(5) Automatic Frequency Control (see RX/F-173-B)

Provision is made for either automatic or manual frequency control of the local oscillator. Selection of one or the other is made by the use of a switch on the front panel.

The A.F.C. input comes from the pre-amplifier as described above, through a 70-ohm concentric line, and drives a tuned circuit in the grid of a conventional 6AC7 I.F. amplifier stage ( $V_1$ ). The output of this amplifier goes to a coil tuned to the mid pass band frequency of the main receiver I.F. amplifier (30 mc/s). Inductively coupled to this coil are two coils, one tuned to approximately 1 mc/s above, and the other to approximately 1 mc/s below this frequency (31 mc/s. and 29 mc/s.).



Each of these coils is coupled to the grid of a section of a 6SN7GT dual triode ( $V_3$ ) through a grid leak and condenser. The plate circuit of  $V_3$  consists of a plate-to-plate resistor load, across which is connected the input of a 6AC7 d.c. amplifier ( $V_4$ ). A polarity reversing switch, and suitable decoupling circuit elements are also included in this circuit.

The action of the discriminator is as follows: Pulses produced by the transmitter beating with the local oscillator are amplified by the I.F. and applied to the input of the discriminator. If the mean frequency of the pulse is lower than 30 mc/s., the coil tuned to 29 mc/s. has a larger induced voltage than the one tuned to 31 mc/s., and conversely. Rectification in the grids of the discriminator tubes produces a bias roughly proportional to the induced voltages in the coils. Since these applied voltages are pulses of short duration, the time constant, RC, the grid leak condenser combination in each grid, is made as long as possible to lengthen the time the d.c. bias, due to each pulse, persists. The output from the discriminator circuit is taken from plate to plate of the discriminator tubes where a voltage appears roughly proportional to the difference in the induced voltage in the two grid coils.

In the plate circuit of the d.c. amplifier, a 0-1 milliammeter (called the A.F.C. Indicator) reads the current through this resistor and hence the output voltage of the amplifier. A 2 mfd condenser connected across the 200 K resistor stabilizes the output and prevents sudden changes in voltages due to small disturbances in the equilibrium of the system. The output of the d.c. amplifier is connected directly to the reflector of the 723-A local oscillator, when A.F.C. is used. When manual tuning is employed, the reflector is connected directly to a potentiometer, and the reflector voltage under these conditions is independent of anything in the A.F.C. circuit.

The 6AC7 d.c. amplifier ( $V_4$ ) utilizes a large amount of cathode degeneration in order to stabilize its operation and reduce the effect of varying tube characteristics. The screen voltage is adjustable in order to set the operating point, or d.c. output of the amplifier with zero input. The voltage from the discriminator tube varies the output voltage around this operation point.

A 6SN7GT multigibrator ( $V_2$ ), having a repetition frequency of from one to two c.p.s. produces, after suitable filtering of its output, an approximately sinusoidal sweep voltage. This voltage is applied to the grid of the d.c. amplifier attenuated to a value sufficient to cause the output of the d.c. amplifier to sweep through approximately 15 volts. The meter referred to above as the A.F.C. Indicator is caused to vibrate slowly at sweep frequency. This sweep voltage is applied to the d.c. amplifier for two reasons:



1. To increase the pull-in range. The local oscillator must by some means be brought close enough to tune to produce a signal into the discriminator, enabling the A.F.C. circuit to pull the local oscillator to the correct frequency and hold it there, hence sweeping the reflector voltage and thereby sweeping the frequency of the local oscillator increases the chances of pull-in occurring.
2. To provide an indication of proper operation of the A.F.C. system. When the local oscillator is being properly held to frequency by the A.F.C. circuit, the system is in a state of equilibrium which it is difficult to upset. The sweep voltage is treated like any other small disturbance and almost completely cancelled by an opposite and equal change in voltage from the discriminator. Thus, proper operation of the A.F.C. circuit is indicated by a cessation of the sweeping of the A.F.C. Indicator.

(6) Power Supply for R.F. Assembly (see RX/F-173-B)

All power in the rack, exclusive of the modulator, is supplied by a 500 c.p.s. line entering the receiver section of the rack through a filter to prevent modulator hash being conveyed into this section. A transformer supplies all filaments and has a 900 volt winding. This winding is used, together with a half-wave rectifier and filtering system, to supply approximately 35 ma. to a load consisting of five VR-150's in series. The various parts of the circuit are connected in such a manner as to apply the proper regulated voltage to each component. A second 250 volt centre-tapped winding supplies, through a full-wave rectifier and RC filter, approximately 100 volts d.c. to be used as the plate supply for the discriminator tube. The latter supply is necessary to permit a single-ended output from push-pull plates.

(7) Pulser (see RX/F-176-B and #661)

PRECAUTION - BEFORE OPERATING THE EQUIPMENT GROUND THE RACK

(a) Introduction - The pulser consists of two artificial transmission lines which are resonantly charged from an 8 KV 13 ma. d.c. power supply, and discharged by means of a rotary spark gap turning at approximately 3720 r.p.m. producing a recurrence frequency of 500 c.p.s. The lines are charged with a 40 henry inductance to approximately 13 to 14 KV and the charge is held at this value by means of the RKR 72 diode until the gap fires.

The functions of the lines are as follows: The lines are initially charged to, say 2E (see fig. 2). When the spark gap fires a -2E unit function propagates down the line (see fig. 3) when it strikes the junction between the two lines, it is looking into an impedance of  $3Z_0$ , i.e. three times its



characteristic impedance. Part of the unit function will be reflected reversed in sign, and part (-E) will pass into the second line (fig. 4). This will produce a voltage  $3E$  between the winding of the second line and ground, and since the load impedance is  $2Z_0$  and the line impedance is  $Z_0$ , approximately  $2E$  will be now developed across the load.

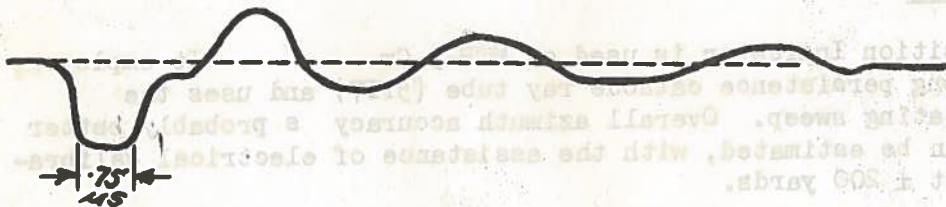
The waves will be reflected as  $-E$  and  $+E$  from the short circuited and the open end respectively (see fig. 5), and they will cancel each other out at the junction of the two lines, at which time the pulse across the load will drop to zero.

(b) Line Description - The lines are each approximately 500 ohm characteristic impedance, five section. Their maximum rating is 15 KV. The lines built at N.R.C. are mica dielectric whereas the Sprague Specialties lines are paper dielectric.

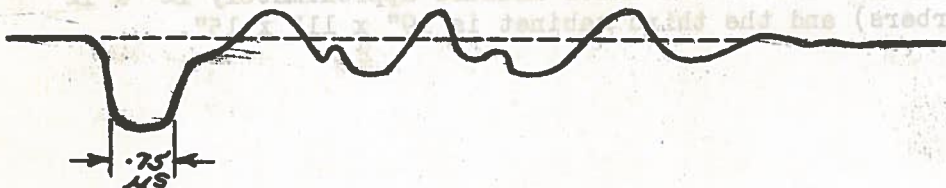
(c) Pulse Shape, Voltage and Current - On a 1000 ohm resistor load the pulse rises in 0.15 microseconds, has a flat top of 0.75 microseconds and falls in 0.27 microseconds.

A high speed oscillograph should be triggered from the spark gap rather than from voltage derived from the output, since the spark gap fires  $3/8$  of a microsecond before voltage appears at the load; this will aid in getting the sweep started before the pulse is applied to the vertical plates.

On a magnetron load the pulse shape is shown below. No.R.F. is put out by the magnetron on any of the past pulse oscillations.



Pulse shape when using paper dielectric network (Sprague Network)



Pulse shape using mica dielectric network (N.R.C. Network)



The mica dielectric network has a lower loss and hence the reason for the slower degree of damping on the past pulse oscillations.

On a 1000 ohm resistor load the peak voltage out should be approximately 11.4 KV at a voltage of 95 volts on the output of the variac, as read on the meter on the control unit.

On a 3 cm. magnetron load such as used in the MTB assembly, the voltage will also be approximately 11.4 KV at 95 volts on the output of the variac. It will increase only a very small amount as the variac is turned up, since the magnetron resistance drops rapidly with a slight increase in voltage.

At 11.4 KV across the magnetron the peak current should be approximately 10 amps.

The power supply current as read by connecting a milliammeter across the 100 ohm resistor in the centre top of the H.T. transformer should be approximately 12 to 13 ma. at 95 volts on the output to the variac at a recurrence rate of 500 cycles. It should not exceed 13 ma.

The spark wheel should be adjusted to run at 3740 r.p.m. by adjusting the rheostat in the control unit when the equipment is initially installed.

With a 500-cycle generator wave form such as used at N.R.C. the correct voltage setting is 90 to 95 volts.

#### V - INDICATOR SYSTEM

##### (1) General Information

A Plan Position Indicator is used on MTB-3 Cm. It employs a 5-inch flat-face, long persistence cathode ray tube (5FP7) and uses the Selsyn type radial rotating sweep. Overall azimuth accuracy is probably better than 1.5° and range can be estimated, with the assistance of electrical calibration rings, to at least  $\pm 200$  yards.

The indicator system follows the "package design". It consists of three units of average weight 50-60 lbs., linked together by plugs and cables. Each unit may be mounted on shock absorbers or all may be installed in a single rack which is shock mounted. Two of the units measure approximately 12" x 12" x 15" (less shock absorbers) and the third cabinet is 19" x 11" x 15".



The primary power requirement is about 250 watts and 115 volts, 500 c.p.s. A short period voltage fluctuation of  $\pm 3$  volts is allowable and the system will accept a long period fluctuation of  $\pm 10$  volts.

With the exception of the PPI unit, all cables issue from the front of the chasses which are easily serviced by removing four screws and withdrawing from the cabinet.

The PPI unit may be placed at some distance from the main equipment and contains all necessary operating control. The cables issue from the rear of this unit to allow a clear field of vision.

A 5-inch monitor c.r. tube (5AP1) with conventional linear scan is available and is used principally for initial tune-up. No provision has been made for IFF but as a compromise, IFF signals might be displayed on this tube.

(2) PPI Time Base Generator, Main Receiver, Calibration and Monitor  
(see RX/F-166-B and RX/F-167-B)

As before mentioned, the PPI system makes use of the Selsyn type rotating sweep. The Selsyn was retained for this service due to its ruggedness, stability and good angular accuracy. However, several limitations appeared in adapting it to a rotary spark gap modulator. These have been satisfactorily overcome for the most part and the final arrangement is considered usable in practice.

Referring to the circuit diagram for this unit, a negative trigger pulse from the modulator causes  $V_1$  to conduct, triggering the multivibrator  $V_2$  and  $V_3$ . A good negative wave is obtained at the plate of  $V_2$  and a positive wave at the plate of  $V_3$ . This arrangement of tubes was chosen to give best stability and least interaction with other circuits.

The negative wave from  $V_2$  is applied to the grid of  $V_4$  (normally conducting) and a very good positive wave of high amplitude is produced in its inductively corrected plate circuit.

The cathode of  $V_4$  is operated below ground merely to obtain a little more plate voltage due to limitations of power transformer at present used.

The brightening pulses for the PPI unit and monitor c.r. tube are taken from the screen of  $V_4$  through separate capacity corrected attenuators. This was required to avoid interference with the monitor c.r. tube intensity when adjusting the intensity correction circuit in the PPI unit.

The PPI time base is produced by creating a linear base of current in the rotor of a Selsyn generator which is mechanically coupled to the rotating antenna. The 3-phase stator is directly connected to a similar stator around the neck of the PPI cathode ray tube.



A radial sweep is produced on the PPI c.r. tube which follows the angular position of the antenna with very good accuracy - usually as good as the manufacturer's claim of  $\pm 1/2^\circ$  - but also somewhat dependent on capacity unbalance of the feed wires between stators and the equality of the 5 K damping resistors across each phase. These items are not critical but should be given consideration. Tube  $V_4$  supplies a good positive square wave to the rotor of the Selsyn generator through cathode follower  $V_5$ , and a current rises in the rotor at a rate depending on the time constant of the Selsyn and the amplitude of the square wave. Only the first part of the exponential current rise is used and is essentially linear except for the first thousand yards. (discussed later). For longer sweeps it was convenient to reduce the amplitude of the square wave and give it a rising characteristic to compensate for the exponential curvature near the end of the sweep, which would normally occur with a flat topped applied voltage wave. This corrected and attenuated wave is produced by the .02 condenser and 100 K variable resistor connected between grid and ground of the cathode follower. Two ranges are provided, 8500 yards and 30,000 yards. Since the short range is a function of the applied voltage wave (the Selsyn time constant being fixed) the 8500 yard range is about the shortest obtainable without using larger tubes and higher voltage supplies. This sweep rate requires a square wave of about 300 volts peak amplitude and the present 110 V Selsyn has so far withstood this transient voltage without failure. The start of the sweep so formed is rather non-linear within the first thousand yards, the sweep rate being  $1/4$  to  $1/2$  the rate of the succeeding ranges. This is principally due to the excessive capacity in the windings of the Selsyn and is minimized by using the cathode follower  $V_5$  as a driving source. Also  $V_5$  is normally cut off and its turn-on characteristic is probably super-imposed. The positive wave from  $V_4$  must also have a certain rise time which contributes to the knee at the start. Capacity effects are the most serious however, and high surge impedance (100 ohms). Telcothene cables strapped together were found best for linking the Selsyn circuits.

Due to the transformer action of the Selsyn and consequent elimination of the d.c. component in the stator on the PPI c.r. tube, the start of the PPI sweep does not occur at the electrical centre of the tube. The centre of rotation is then some distance along the sweep from its start.

Several methods were tried to achieve centering and many were found to jitter on spark triggering because of the non-uniform repetition periods. The best arrangement consists of forcing a reverse current through the Selsyn immediately following the sweep and adjusting its amplitude or duration so that the average of the wave so formed passed through the start of the sweep. The reverse current had to return to zero well before the next sweep, or else jitter took place. This is accomplished by feeding the grid of  $V_8$  with the positive output from  $V_3$ . Being already saturated by 500 K to +300,  $V_8$  grid does not respond but on  $V_3$  plate returning to normal,  $V_8$  grid is run highly negative for a time determined by 500 K grid resistor and .002 coupling condenser. Thus a



positive square wave of controllable duration is formed at the plate of  $V_8$  and inverted in  $V_9$  which is connected to the Selsyn and causes a reverse current to flow after the sweep wave.  $V_9$  develops cut-off bias due to grid current and operates alternately with  $V_5$ .

With the Selsyn directly connected to ground, the centering required changing for each sweep and was somewhat critical. By inserting a series condenser (2 mfd) centering was automatic and the grid circuit of  $V_8$  became non-critical. This is apparently due to the fact that an increased current through  $V_5$  charges the condenser to a higher voltage and when  $V_{10}$  comes into action the condenser voltage is additive to its supply voltage and an increase reverse current results. This maintains the proper balance and the average of the current wave in the Selsyn remains the same. The beam will remain centered over a considerable range of adjustment of the grid circuit of  $V_8$ .

By connecting a 10 K resistance across the Selsyn coupling condenser by means of a switch, the start of the sweep can be caused to rotate around a small circle for better close-in azimuth accuracy. A slight reduction of sweep speed occurs but is not harmful.

The main receiver and monitor circuits consist of a three stage 30 mc. inverse feed-back I.F. and detector coupled to a single video amplifier,  $V_{11}$ , with a very low plate resistor. The output of this stage is connected by cable to a remote video in the PPI unit and also to a monitor video  $V_{12}$  to display the signals on the monitor c.r. tube. No attempt has been made to inductively correct any of these stages. A narrow pass band has been found quite adequate for PPI and there are some indications that the signal noise ratio is improved thereby.

A fairly linear monitor sweep is produced by inserting a 50 ohm resistor shunted by a .1 condenser in series with the Selsyn rotor. The resultant voltage is amplified and inverted by  $V_{10}$  and is used to provide horizontal deflection of the monitor cathode ray tube. The sweep is a good replica of the current in the Selsyn (allowing for slight distortion introduced by  $V_{10}$ ).

The range calibrator is a conventional shock oscillator started by cutting off the current in  $V_6$  and whose output is turned into a square wave by  $V_7$ . The output amplitude of  $V_7$  is controlled by varying the screen voltage and feeds through a small condenser (25 mmf) into the low impedance input of the first video amplifier. Differentiation occurs and pips are produced which mix with the incoming signals. The pip interval on the short sweep is 1000 yards and on the long sweep the interval becomes 5000 yards. This allows fairly accurate interpolation without the confusion which would exist with a larger number of calibration rings.

### (3) PPI Unit (see RX/F-167-B)

The PPI unit consists of a 5FP7 long persistence c.r. tube with adjustable focus coil for beam centering and using a Selsyn stator around the neck of the c.r. tube for the PPI sweep. All components are rigidly mounted and easily serviced.



A single 1852 video stage is contained in this unit and feeds negative signals into the cathode of the c.r. tube. The brightening pulse is applied to the grid circuit. Provision is made to even the picture illumination by sloping the brightening pulse by means of the variable 100 K resistor and .0025 condenser from grid to ground. A 6H6 double diode combines the function of limiter and d.c. restorer in the signal circuit.

The operating controls on this unit consist of receiver, range switch, calibration intensity c.r. tube intensity and focus. Other controls used less often were also included. These are limiter, cen re shift and intensity correction. The latter may well be screwdriver adjustments in later design.

Provision is made for a Selsyn or magslip to be driven from the cursor for remote azimuth indication.

The 5FP7 is operated with the second anode "hot" to facilitate wiring and to eliminate corona noise from the grid circuit. No trouble has been experienced with this arrangement other than a slight picture distortion for a few minutes when first switched on.

It might be noted here that a stabilized picture referred to true north instead of ship's head might be accomplished by connecting a device such as a differential Selsyn in the leads linking the Selsyn stators. The shaft of the differential Selsyn is directly driven by a gyro compass. Probably extra drive will be required from the sweep generator to make up for impedance mismatch. Also it is thought improbable that a sweep shorter than 15000 - 20000 yards will be obtained.

#### (4) Power Supply (see RX/F-163-A)

The power supply for the indicator system is quite straight-forward. A 5600 volt R.M.S. 5 ma. transformer with 2.5 V 5 A rectifier winding supplies the requirements of the c.r. tubes. The rectifier is a 2V3G.

The low voltages are obtained from a transformer with one winding of 375-0-375 volts RMS, 200 ma and another of 275-0-275 volts RMS 70 ma. A 6.3 V. 10 ampere filament winding supplies all heaters. The rectifiers use type 5U4G which require 5 volts, 3 amperes each.

Condensers are all oil impregnated paper and the .1 mfd 7500 V types may be reduced to .03 mfd. No smaller sizes were available for the first design.

Three VR-150 tubes are connected in series (from +350 to -100) to provide a regulated voltage for  $V_4$ . This amount of regulation was found sufficient to keep the sweep speed fairly constant for a considerable change in supply voltage.



A special filament transformer wound on a Hammond 270 core with a cross section of 1.5 square inches isolates heaters of  $V_4$ ,  $V_5$  and  $V_9$ . It consists of 72 turns of #16 D.C.C. on the primary and three separate secondaries, each of 85 turns #22 D.C.C. Insulation is rated at 500 volts d.c.

## VI - OPERATING INSTRUCTIONS

### A. NORMAL TURN-ON INSTRUCTION

- (1) Switch on a.c. and d.c. by interlocked switches on the control box. This supplies power to all units except the modulator high voltage and the d.c. for the spinner. The monitor sweep should be in operation at this point, but running at a slow repetition rate.
- (2) The high voltage is turned up slowly on the modulator to a final value of about 95 volts, as read on the control box meter.
- (3) The monitor sweep should now be synchronized with the transmitter pulse and its intensity and focus adjusted by the lower left- and right-hand controls respectively, in the large service port on this chassis. Centering is obtained by means of the upper two knobs in the small service port.
- (4) Make sure the "remote-near" control switch is in the down position, and the receiver gain (centre left-hand knob) may be turned up at this time. The transmitter pulse will be seen at the left edge of the sweep, and customary receiver noise will fill in the balance.
- (5) The "Range-change" switch should now be operated to give the widest transmitter pulse -- this will give a sweep length of about 7500 yards. This switch must not be held in, as only a momentary contact is necessary to operate the range relay. Should this circuit be inoperative, examine the 1 A. relay fuse (just to the left of the push-button).
- (6) The amplitude of the calibration pips may be adjusted by means of the right-hand centre knob. The calibration interval is 1000 yards on the 7500-yard sweep, and 5000 yards on the 30,000-yard sweep.
- (7) The sweep is adjusted to give the 30,000-yard range by means of the pre-set control in the small service port.
- (8) Operation Instructions for the R.F. Chassis
  - (a) After the power is turned on at the control box, crystal current should be observed as soon as the local oscillator tube has warmed up (usually about one minute). If crystal current is not obtained, check the oscillator current (switch  $S_2$ ), and if it is satisfactory (between 15 and 25 ma.) (vary potentiometer P, and crystal current will appear. Adjust the local oscillator coupling until the crystal current is about 0.4 ma. or so.



- (b) With  $S_1$  set at "manual" and the spinner off, turn on the magnetron by the variac control on the control box until the magnetron current is 3 - 4 ma. Check the R.F. output to the antenna with a neon bulb.
- (c) Point the antenna in a direction to give an echo. Echoes should appear on the monitor sweep as the local oscillator is tuned manually, either by potentiometer P, or by the mechanical screw on the resonator of the tube. With a new 723-A tube, both methods will probably be needed, but normally only the potentiometer P is needed when turning on the set.
- (d) To adjust the crystal matcher, decouple the local oscillator until the crystal current is 0.1 ma. or less. If this is not done the matcher pulls the local oscillator frequency and the proper match will not be obtained unless the local oscillator is retuned at each step in the matching. With low crystal current, adjust the two matching variables (length and position of the screw) in turn, until some selected steady echo reaches a peak. Lock the position adjuster on the matcher and couple up the local oscillator to increase the crystal current until the best signal-to-noise ratio is produced on the monitor sweep. The best crystal current is usually 0.3 to 0.6 ma., depending on the crystal.
- (e) Tune the T/R and R/T for maximum signal in each case, and lock the tuning slugs.
- (f) To transfer to A.F.C., adjust the potentiometer P for maximum signal and throw  $S_1$  to the up position. The A.F.C. indicating meter will now oscillate due to the sweep, and if the correct tuning voltage is within the voltage sweep, the circuit will lock in, the oscillations will cease, and the A.F.C. is operating correctly. If this does not happen, adjust potentiometer P, which varies the mid-point of the voltage sweep, until either the "notch effect" or "wall effect" is observed. (See A.F.C. Section 8). The "notch effect" means proper A.F.C. control, and the "wall effect" means that the wrong side band is being tuned. If the "wall effect" is observed, throw  $S_3$  to the other position - this changes the polarity of the A.F.C. output voltage (other side band!), and now if potentiometer P is again varied through its range, the A.F.C. will find and control the frequency of the local oscillator.

The crystal matching may be done easily on the A.F.C. with high local oscillator coupling, since the A.F.C. will automatically correct for the pulling of the local oscillator by the matcher.

**NOTE:** - Cessation of the oscillation of the A.F.C. indicating meter means proper A.F.C. operation. If the oscillation does not stop under the above conditions, the circuit is out of adjustment. (See alignment of A.F.C. circuit).



- (9) Assuming the R.F. portion has been adjusted to give proper operation (see R.F. instructions), the "remote-near" switch is thrown to the up position, and the set is now operative from the PPI unit.

The controls on the PPI, starting from the left of the bottom row are: calibration intensity, receiver gain, intensity and focus.

In the second row, similarly, we have: centre shift switch (rotates), limiter threshold, intensity (picture illumination) correction.

The range switch (momentary push) is separate, and to the lower left of the PPI part.

The beam centering knobs are near the left and right upper corners.

(10) To adjust the PPI

- (a) Centre the sweep - this requires some care or azimuth accuracy will suffer. Set the spinner turning and adjust the centering controls until the centre of the beam rotation is coincident with the cursor centre. This will give fair accuracy. For best results check the beam for coincidence at two points exactly 90° apart on the azimuth indicator, and make any necessary re-adjustments to the centering controls.
- (b) Adjust the intensity so the beam is barely visible.
- (c) Set the R.F. portion in operation and turn up the receiver gain (with spinner stopped). A series of bright dots corresponding to any echoes present will appear on the PPI, with noise appearing as random fluctuations of intensity.
- (d) The limiter and focus controls should be adjusted for best definition. In the absence of echoes, the calibration intensity may be turned up and a similar procedure used.
- (e) Set the spinner rotating and a plan indication of all targets in the area should be visible. The azimuth is estimated by splitting the echo with the cursor, and range is estimated by turning up the calibration intensity and interpolating between the rings.
- (f) For better azimuth accuracy on close-in targets, the centre shift switch is operated. This allows the centre to rotate about a small circle.
- (g) To even the picture illumination, the intensity correction knob is adjusted to reduce the beam intensity at the start. The intensity control should be adjusted at the same time to give best results. Ultimately this may be a pre-set adjustment.



In operation receiver gain, range, calibration intensity and intensity controls will be used frequently. The remainder of the controls will seldom be adjusted.

B. LINING-UP A.F.C.

Adjustment of A.F.C. Circuit

To adjust the A.F.C. circuit first align the d.c. amplifier portion as follows. Remove the 6SN7GT sweep tube ( $V_2$ ) and turn on the set, allowing two minutes at least for the tube characteristics to stabilize. With the A.F.C. indicator reading approximately in the centre of its scale, flip the side band switch ( $S_3$ ). If the A.F.C. indicator changes reading, adjust the 2 K potentiometer inside the chassis (the one nearest the back) until the A.F.C. indicator reading is unaffected by throwing the side-band switch ( $S_3$ ).

With the switch,  $S_1$ , in the "manual" position, turn the potentiometer P, on the control panel fully clockwise. Determine the voltage on the 723-A reflector thus applied, by pushing  $S_4$  on the control panel, and noting the reading on the A.F.C. indicator. The reading should be approximately full-scale, or 200 volts. Throw  $S_1$  to the A.F.C. position, and adjust the 3 K potentiometer inside the chassis until the meter reading is the same as for "manual". Turn the potentiometer P, on the control panel, fully counter-clockwise, and again make the readings on the A.F.C. indicator coincide on A.F.C. and on manual; this time by adjusting the 15 K potentiometer inside the chassis. The A.F.C. indicator reading in this case should be approximately 0.4, or 80 volts. Replace the 6SN7GT sweep tube.

The above procedure should be run through whenever the 6SN7GT discriminator tube ( $V_3$ ) or the 6AC7 d.c. amplifier ( $V_4$ ) is replaced.

To ascertain if the discriminator part of the A.F.C. is operating properly, connect the output of a G.R. 804B signal generator in the crystal socket. This connection may conveniently be made by soldering a spike from an old crystal to a Jones plug tip. Now remove the 6SN7GT sweep tube  $V_2$ . With the switch on A.F.C., adjust the A.F.C. indicator to a mid-scale (100 volts). Sweep the frequency of the signal generator back and forth from 27 - 33 mc., and adjust the input voltage until a maximum swing of 20 volts in either direction is obtained. If the discriminator is properly aligned, the A.F.C. indicating voltmeter will read 80 volts at 29 mc.; 120 volts at 31 mc. (or vice versa), and 100 volts at 30 mc. This swing should be produced at a reading of approximately 150 microvolts on the output of the signal generator attenuator. If the swing on the A.F.C. indicating voltmeter, produced by sweeping the frequency from 29 - 31 mc. is not symmetrical about 100 volts, or if the reading on the output attenuator of the signal generator is much different from the figure of 150 microvolts, then the A.F.C. is out of alignment and must be aligned as follows:



- (1) Set the signal generator at 30 mc. and increase the signal until a deflection is produced on the A.F.C. indicating voltmeter.
- (2) Tune the centre coil of the discriminator, the input coil to the I.F. amplifier ( $V_1$ ) driving the discriminator, and the coils in the second and third stages of the pre-amplifier for maximum deflection of the A.F.C. indicator.
- (3) Set the signal generator at 29 mc. and tune the side discriminator coil which has the greatest number of turns to maximum deflector up or down on the A.F.C. indicator.
- (4) Re-set the signal generator to 31 mc. and tune the other side coil for maximum deflection down or up, on the A.F.C. indicator.
- (5) Check for symmetry of the deflections for 29 mc. and 31 mc. from the 30 mc. reading. The side coils are mounted in slots in the bottom of the chassis and their position relative to the central coil must be adjusted to achieve equality of the two deflections.

The sensitivity depends to a considerable extent on the spacing of the discriminator coils. If the gain is greatly different (say 100%) from that specified above, (20 volts deflection for 150 microvolts input) it may be corrected by adjusting the positions of the two side discriminator coils; the balance of course, should be maintained. If the correct gains are not easily obtained this way, then the gain of the pre-amplifier, and the I.F. driver stage on the A.F.C. chassis should be checked.

- (6) Remove the signal generator; insert the crystal and completely de-couple the local oscillator. Turn up the modulator until the magnetron draws normal pulse current. The A.F.C. indicating meter reading should not change as the modulator is turned up. If the meter tends to deflect from the central position, tune the input coil on the pre-amplifier until the deflection is a minimum.
- (7) Insert the sweep tube 6SN7GT ( $V_2$ ). The action of this circuit should produce about 15 volts periodic swing on the A.F.C. indicator with a period of about 1 second.

The tuning of the output coil of the pre-amplifier (third stage,  $V_3$ ) should be done with a modulated signal generator, and the tuning of the first stage input coils should be checked on an echo. It should of course, be the same as that found previously by checking the influence of the transmitter pulse on the A.F.C. circuit. (See (6) above).



(8) The A.F.C. is now ready to operate. Obtain a signal with manual tuning and throw the A.F.C. switch. If the A.F.C. does not lock in, adjust the potentiometer P, since the sweep voltage may not be sufficient to reach the proper frequency. Two things may happen:

(a) "The notch effect" - the proper operation of the A.F.C.

(b) "The wall effect" - As the potentiometer is increased or decreased, the meter needle will swing up or down to a certain reading and no farther. At the same time the periodic sweep will be diminished and almost eliminated. This effect is caused by having the local oscillator operating 30 mc. above the magnetron; whereas it should be 30 mc. below (or vice versa). This may be immediately corrected by throwing switch S<sub>3</sub>, which reverses the polarity of the input to the d.c. amplifier from the discriminator.



FIG. 1.

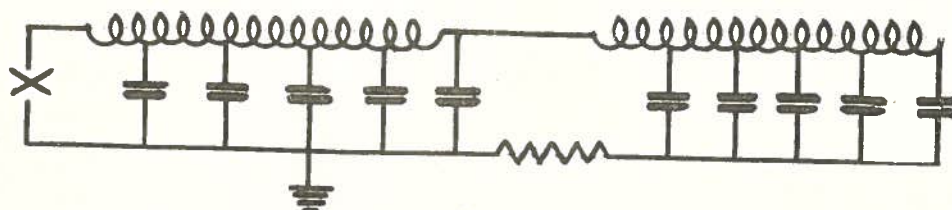


FIG. 2.

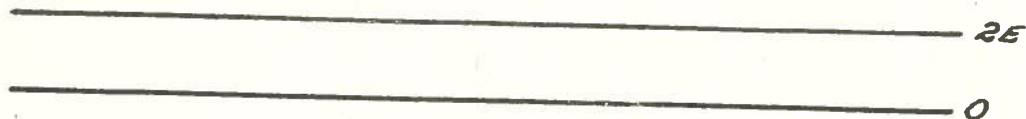


FIG. 3.

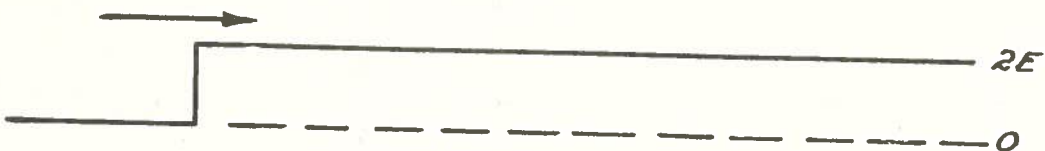


FIG. 4.

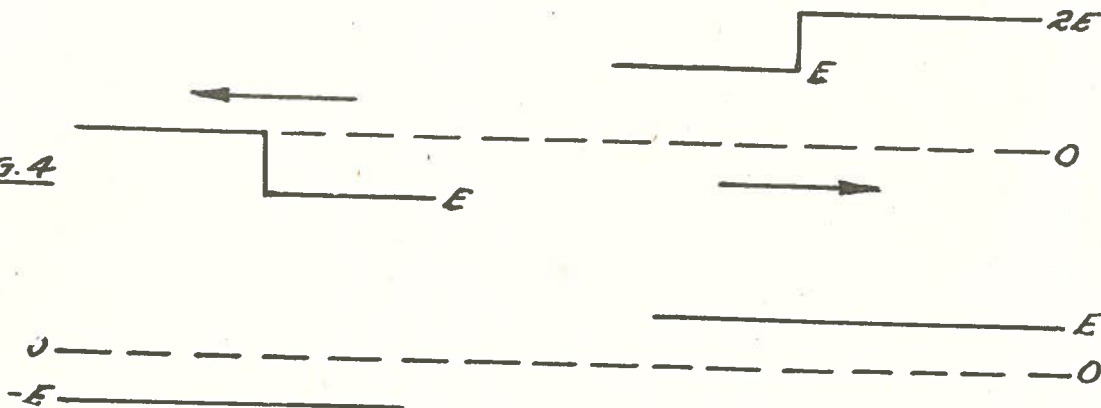
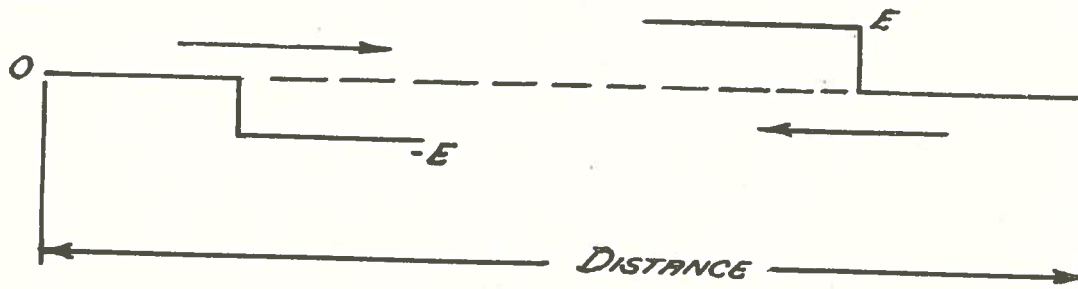


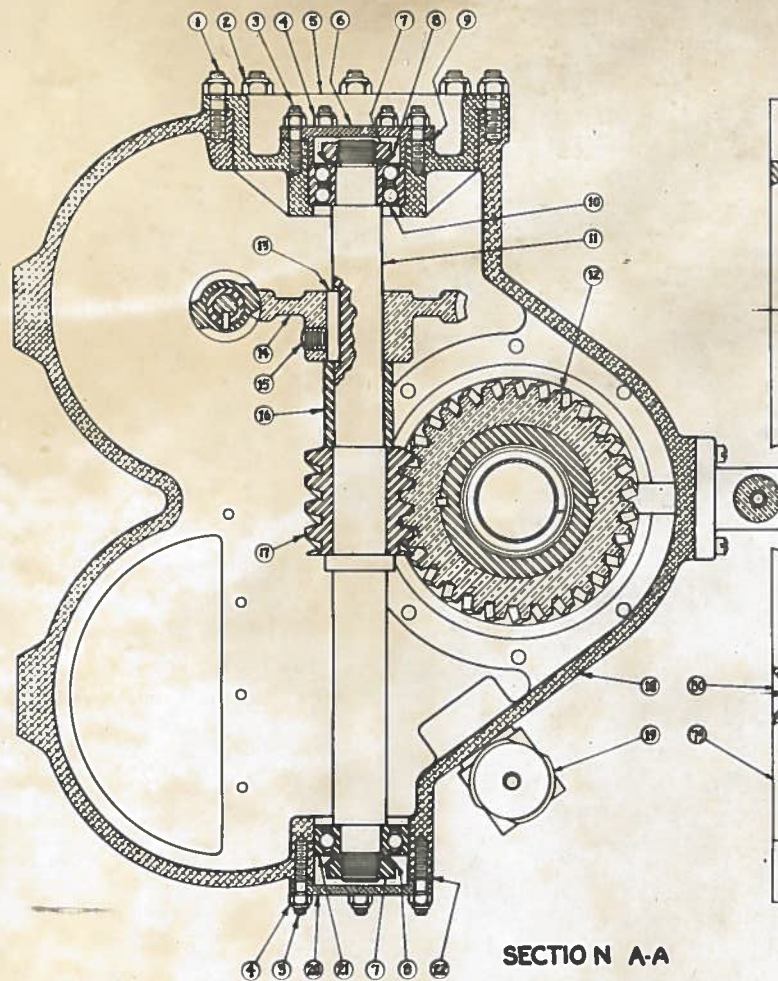
FIG. 5.



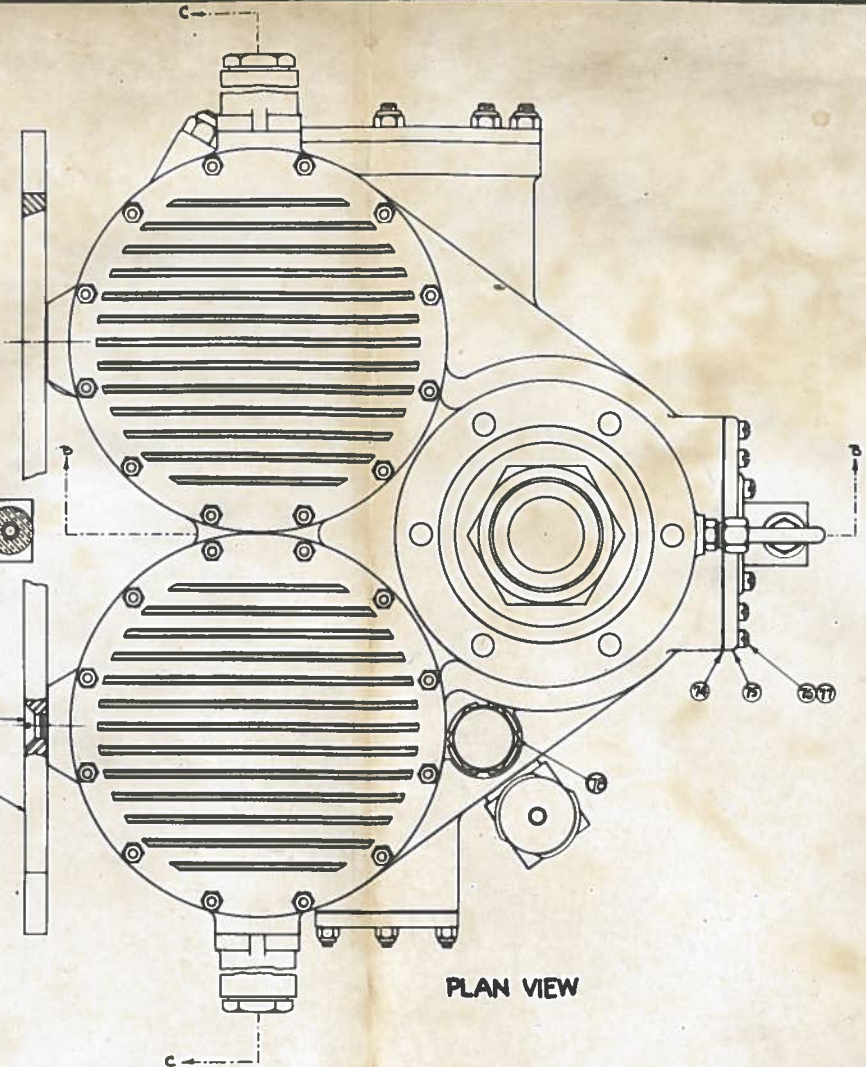
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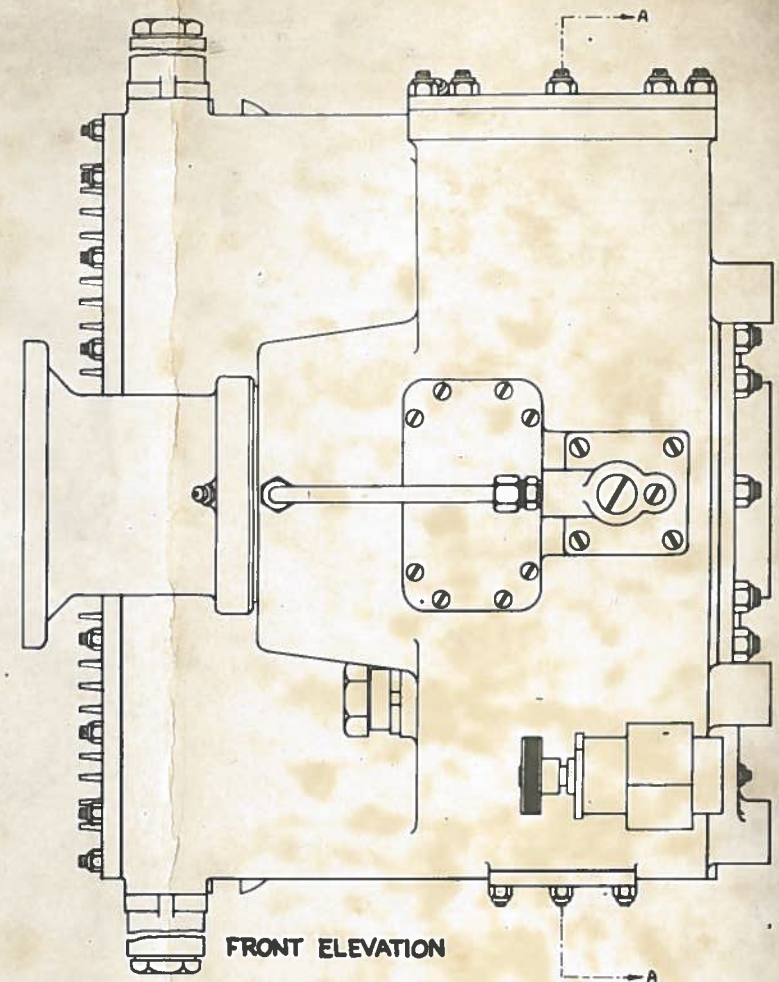
NO.	NAME
1	STUD
2	W/3 ELASTIC STOP NUT
3	STUD
4	W/3 ELASTIC STOP NUT
5	THROAT BRG. FLANGE
6	COVER
7	W/3 LOCK NUT
8	W/3 LOCK WASHER
9	GASKET
10	HD BRG 5203
11	COUNTER SHAFT
12	WORM WHEEL OF 4 START
13	W/3 KEY
14	WORM WHEEL OF 2 START
15	BRG. SOCKET SCREW
16	SLEEVE SPACER
17	SP. 4 START WORM
18	HOUSING
19	COVER COCK ASSY
20	COVER
21	HD BRG 3203
22	GASKET
23	COVER
24	SELYN 2JDSF2
25	CLAMPING RING
26	SELYN MOUNTING
27	SELYN MOUNTING CLAMP
28	4-40S ALG. BRG. SCREW
29	SELYN CLAMP
30	HD BRG 55504
31	COVER
32	KEYED SLEEVE
33	LOCK WASHER
34	SELYN SHAFT SCREW
35	4-40S ALG. BRG. SCREW
36	GASKET
37	SELYN SPROCKET
38	BRG. AIRCRAFT CHAIN
39	STEEL LOCK WASHER
40	HD 32# BRG. SCREW
41	GASKET
42	GASKET
43	W/3 GLAND NUT
44	PACKING NUT
45	D.C. MOTOR
46	TAPER PIN CLAMP
47	W/3 ELASTIC STOP NUT
48	END BELL
49	GEAR END 10 M. D
50	W/3 WASH. KEY
51	END BELL SCREW
52	BUSHING
53	12 P. WORM WHEEL
54	W/3 WASHER
55	RETAINING SCREW
56	GASKET
57	ON SEAL TUBE
58	CHUCK COUPLING
59	SUPPORTING FLANGE
60	3/4" X 3/4" WOOD RUFF KEY
61	"BRK" FITTING ON
62	GAS LOCK KLOUSE ON
63	3/4" X 3/4" WOOD RUFF KEY
64	ECCENTRIC SPACER
65	ON VIB. ASSY
66	BEARING MOUNTING
67	DRIVEN ON SHAFT
68	SHAFT
69	MAIN SHAFT
70	3/4" X 3/4" W/3 KEY
71	4-40S ALG. BRG. SCREW
72	4-40S ALG. BRG. SCREW
73	TAPER PIN
74	GASKET
75	PROXY INSPECTION COVER
76	3/4" X 3/4" W/3 KEY
77	LOCK WASHER FOR 3/4"
78	ON FILM
79	ON MOUNTING PLATE
80	W/3 W/3 ALG. BRG. SCREW



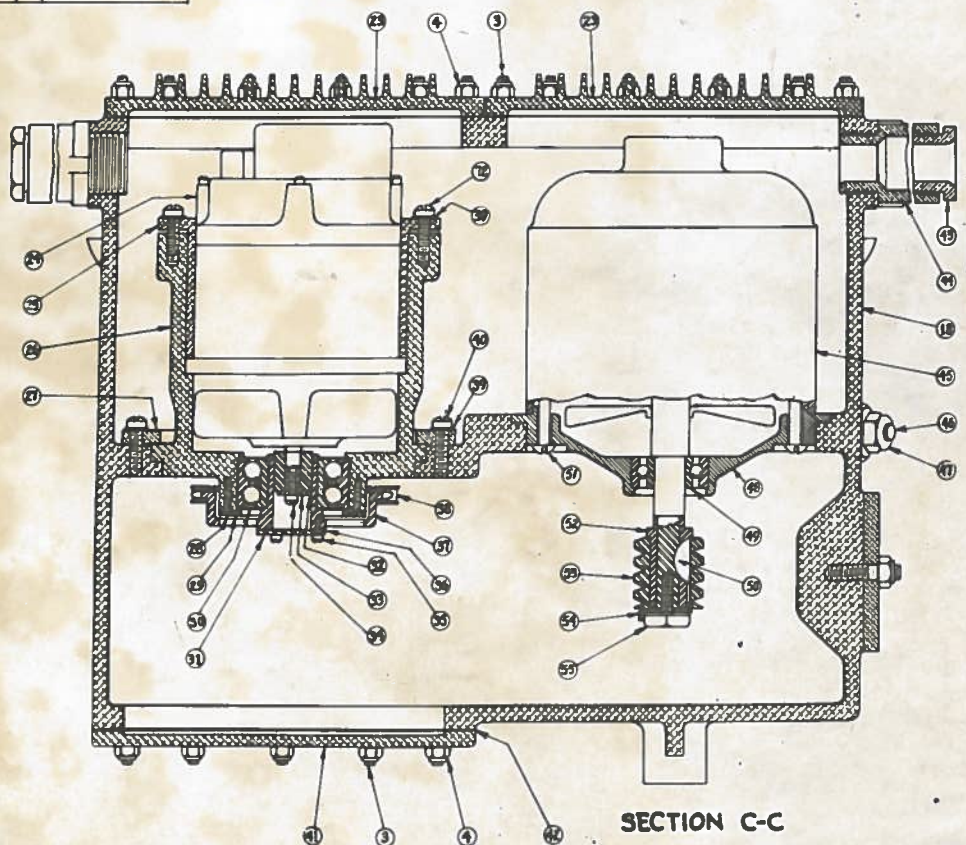
SECTION N A-A



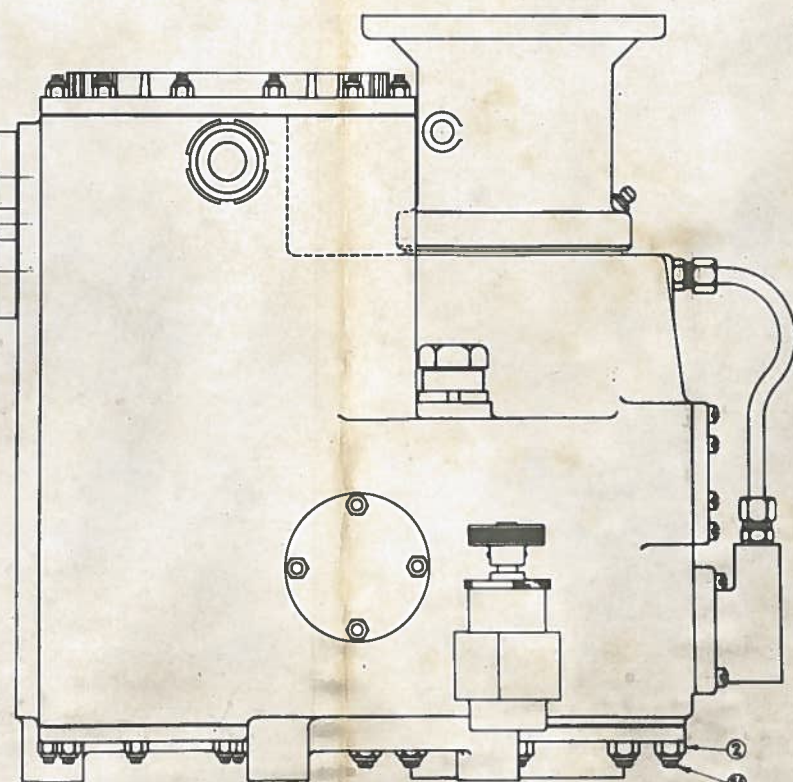
PLAN VIEW



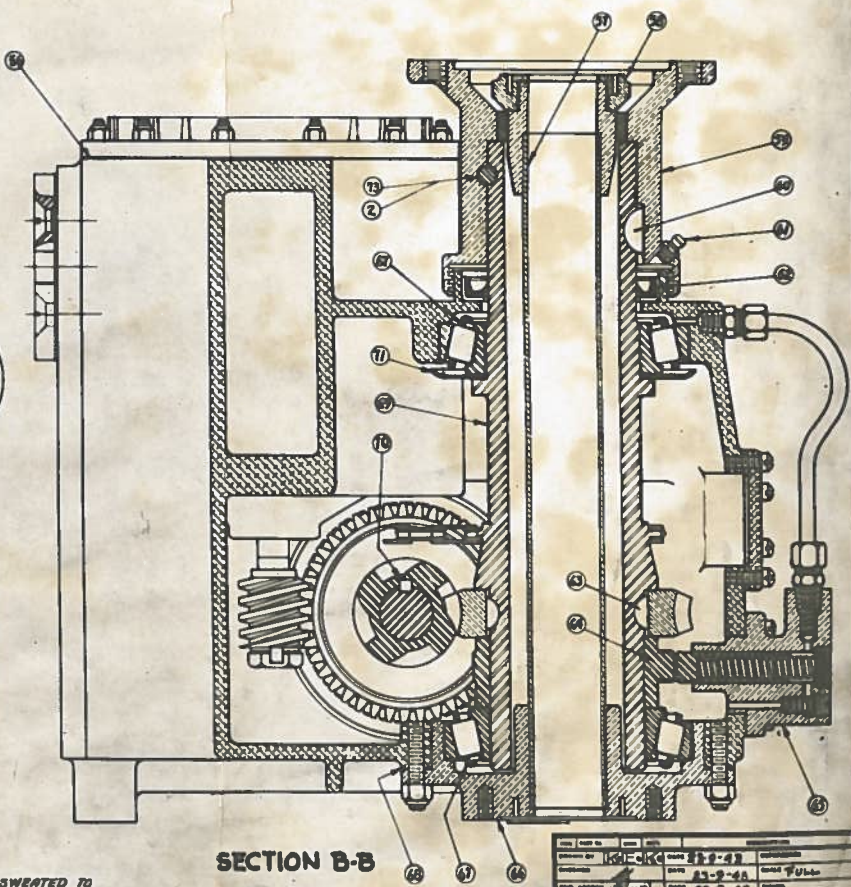
FRONT ELEVATION



SECTION C-C



SIDE ELEVATION



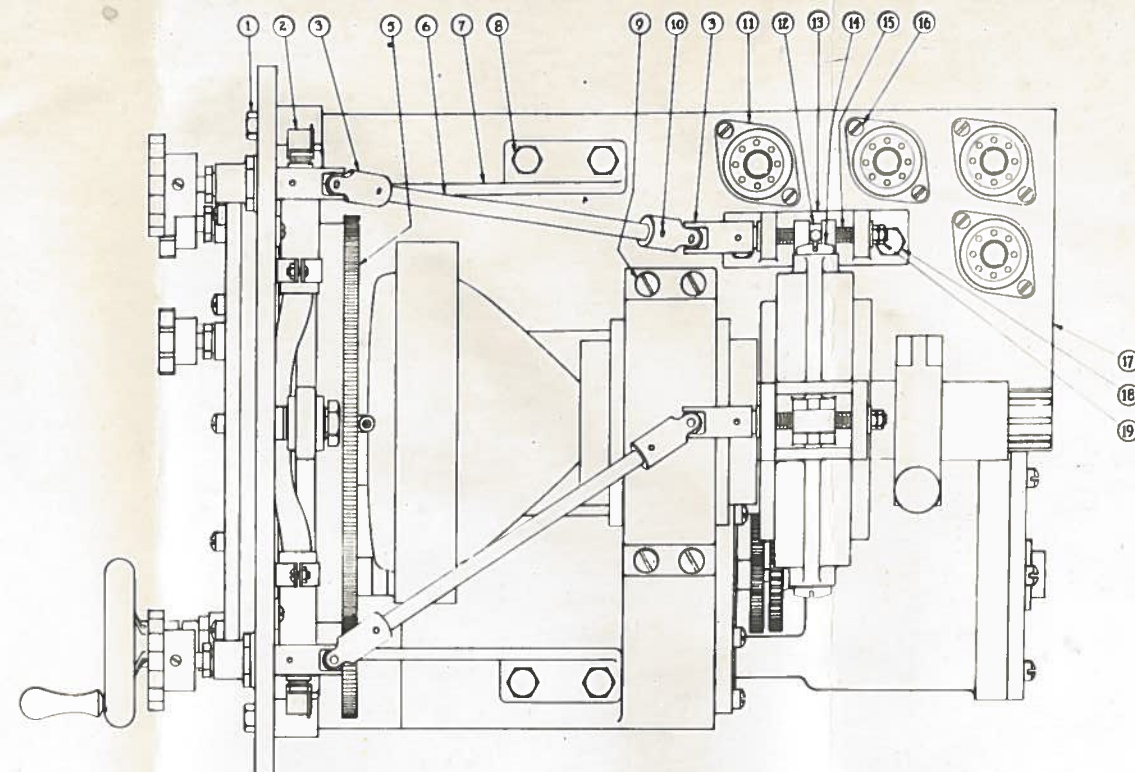
SECTION B-B

REVISED 14-11-43  
PART 27 SWEATED IN  
PART 44 ON ASSEMBLY

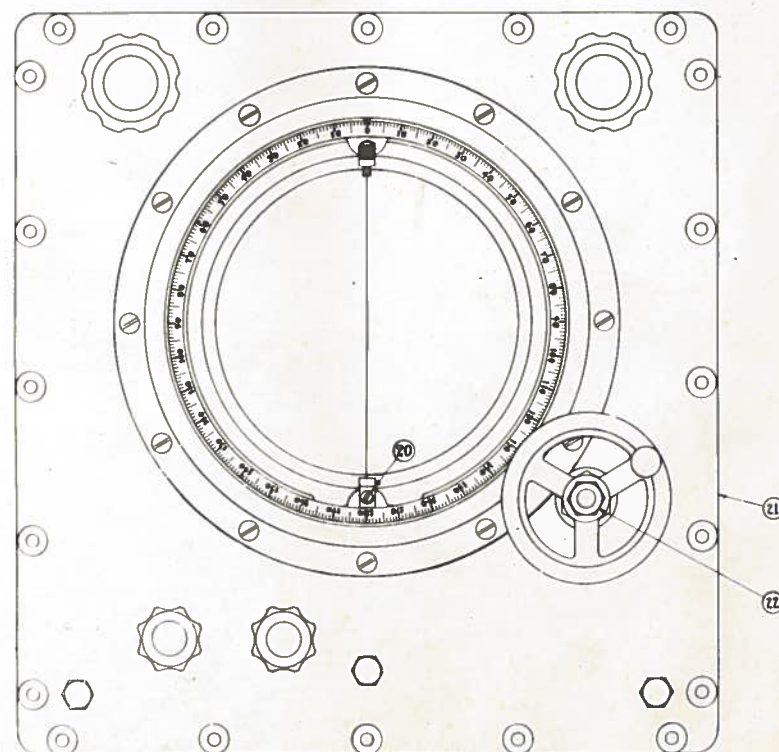
REVISED 14-11-43  
REVISED 12-11-43



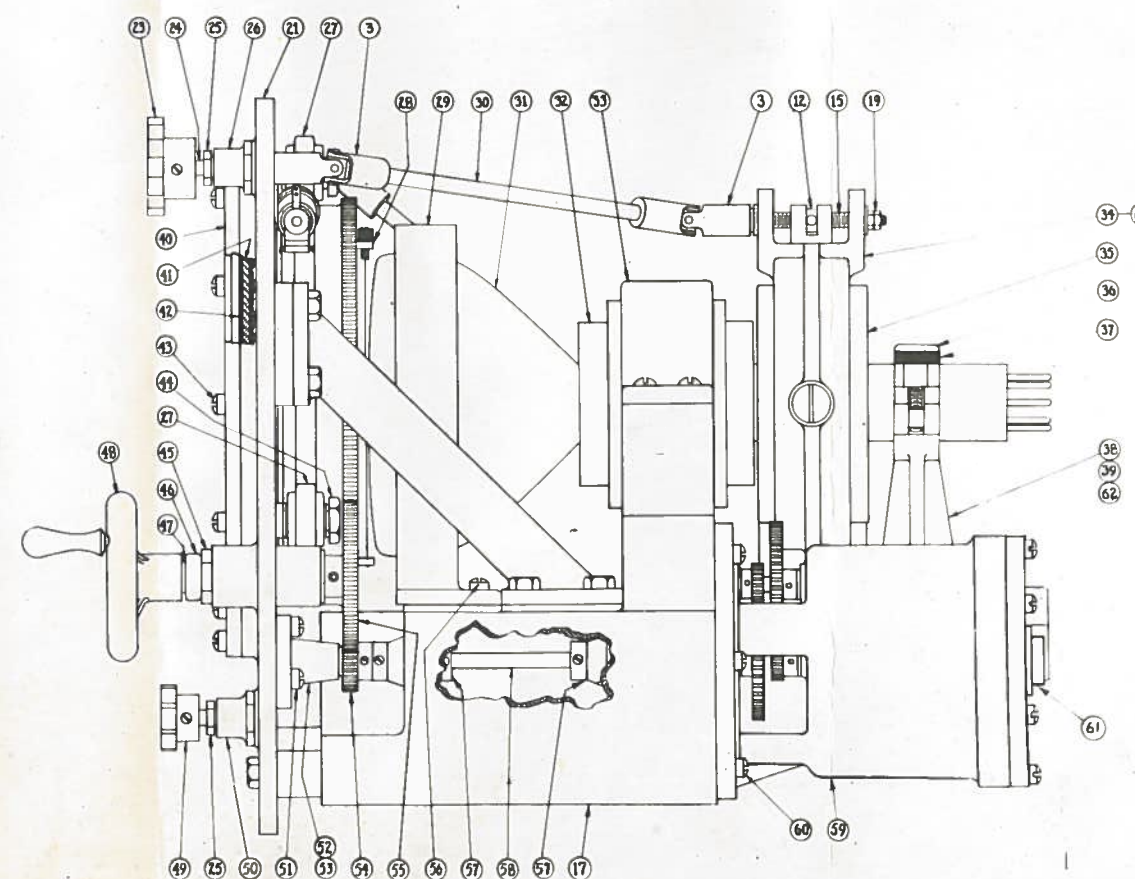
REV	BY	DATE	TITLE
1	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
2	115 D	1	SCALE ILLUMINATION
3	115 D	1	BOSTON UNIVERSAL JOINT C-642B
4	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
5	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
6	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
7	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
8	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
9	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
10	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
11	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
12	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
13	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
14	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
15	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
16	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
17	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
18	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
19	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
20	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
21	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
22	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
23	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
24	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
25	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
26	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
27	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
28	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
29	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
30	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
31	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
32	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
33	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
34	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
35	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
36	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
37	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
38	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
39	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
40	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
41	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
42	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
43	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
44	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
45	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
46	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
47	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
48	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
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53	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
54	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
55	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
56	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
57	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
58	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
59	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
60	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
61	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
62	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG
63	115 D	1	W 20 HEX HEAD M.S. 1 1/2" LONG



PLAN VIEW



FRONT ELEVATION



SIDE ELEVATION

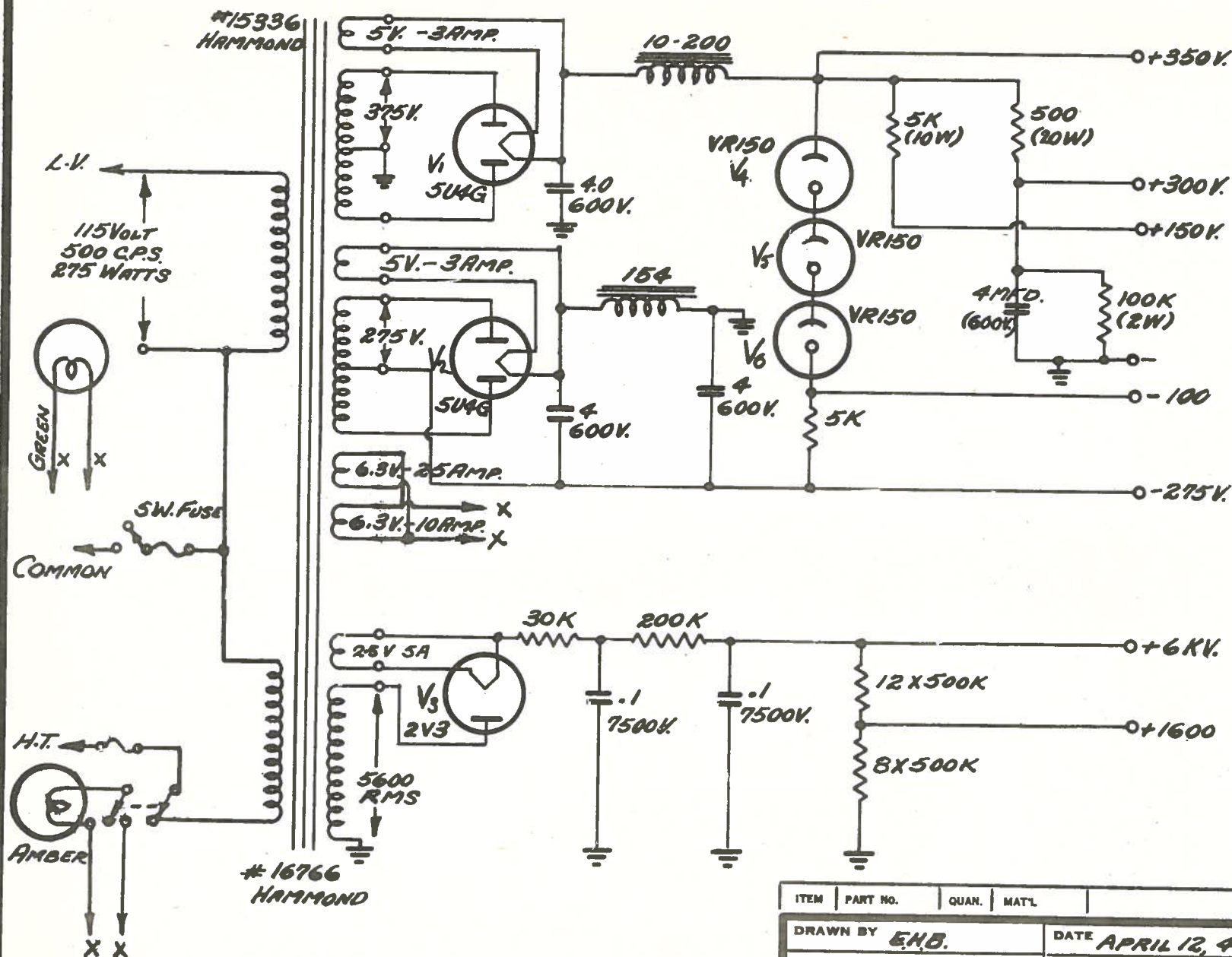
SELSYN CHANGED TO M MOTOR  
AND 6 CONTROLS REMOVED  
REV 28-9-43

28 REVISED - K.E.K. - 15-5-43

30 REVISED - K.E.K. - 15-5-43

DATE	28-9-43	BY	K.E.K.
REVISED	15-5-43	BY	K.E.K.
NATIONAL RESEARCH COUNCIL-RADIO SECTION			
WATERPROOF PPI ASSEMBLY			

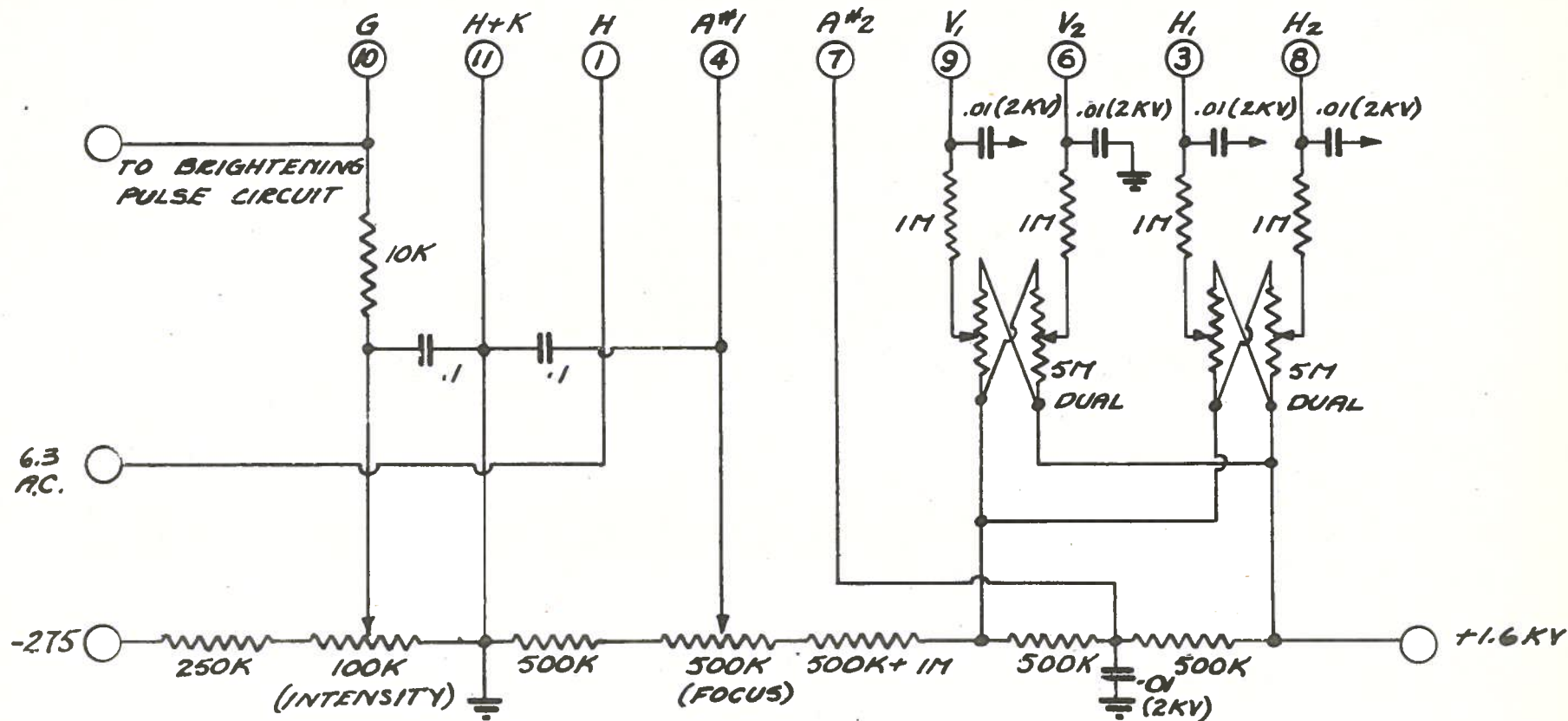




NOTE: VALUES MAY ALTER  
IN FINAL MODEL.

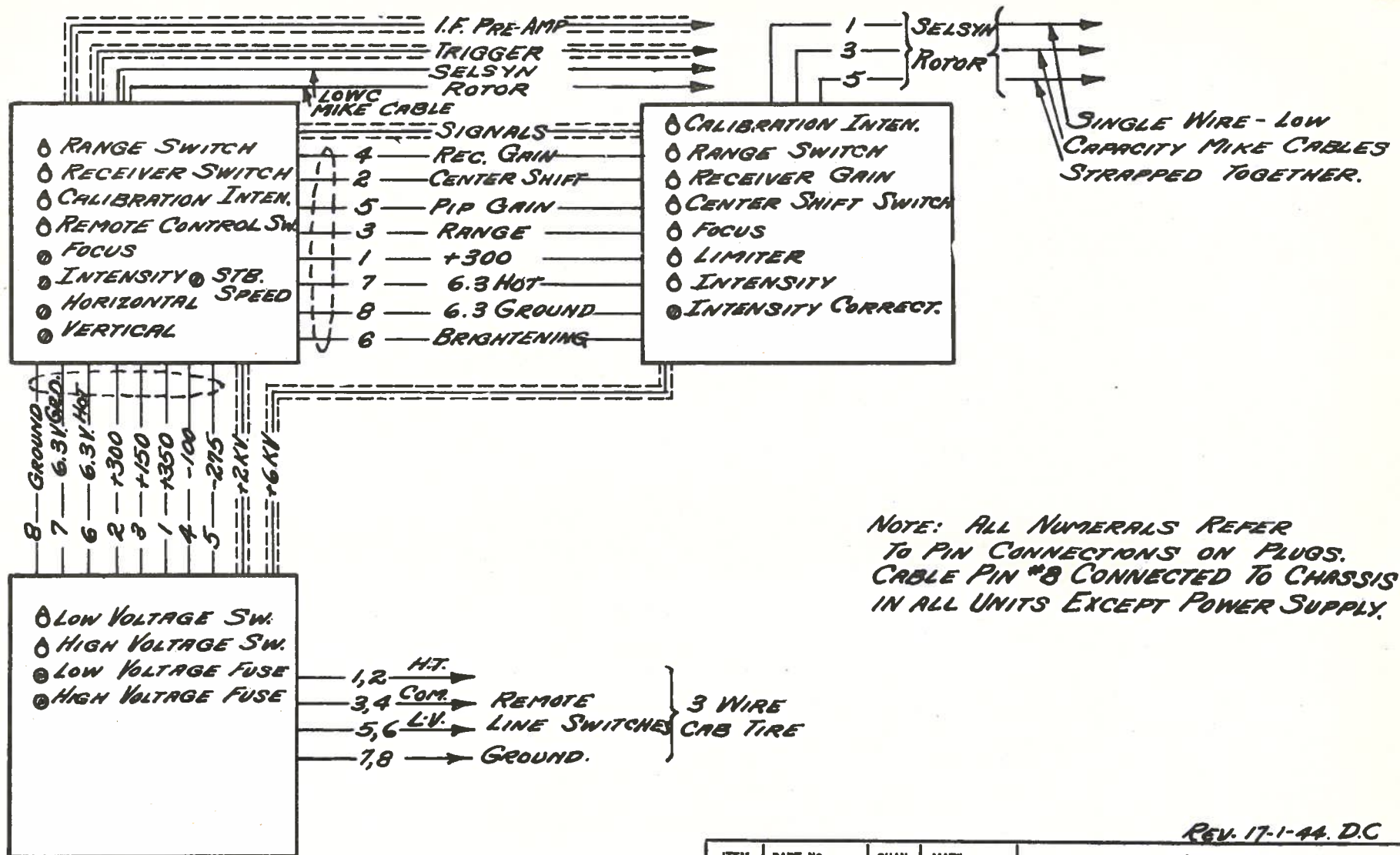
REV. 17-1-44 RC.				DESCRIPTION	
ITEM	PART NO.	QUAN.	MAT'L		
DRAWN BY <i>E.H.B.</i>		DATE <i>APRIL 12, 43</i>		SUPERSEDES	
CHECKED <i>H.W.M.</i>		DATE		SCALE	
ENG. APPROV. <i>H.W.M.</i>		DATE <i>16-4-43</i>		FINISH.	
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA					
NAME <i>CIRCUITS CONSTANTS OF POWER UNIT FOR RX/F INDICATOR CHASSIS TYPE #2</i>				DWG. NO. <i>D-P5</i> <i>RX/F-163-A.</i>	





ITEM	PART NO.	QUAN.	MAT'L	DESCRIPTION
DRAWN BY <i>D.L.S</i>		DATE <i>14-4-43</i>		SUPERSEDES
CHECKED <i>H.W.M.</i>		DATE		SCALE
ENG. APPROV. <i>H.W.M.</i>		DATE <i>16-4-43</i>		FINISH.
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA				
NAME <i>RX/F MONITOR C.R.T. CIRCUIT</i>				DWG. NO. <i>D-CR</i>
				<i>RX/F-164-A</i>





ITEM		PART NO.	QUAN.	MAT'L	DESCRIPTION
DRAWN BY <i>ENB.</i>		DATE <i>APRIL 10, 1949</i>			SUPERSEDES
CHECKED <i>H.W.M.</i>		DATE			SCALE
ENG. APPROV. <i>H.W.M.</i>		DATE <i>16-4-43</i>			FINISH.
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA					
NAME <i>CHASSIS INTERCONNECTIONS</i>				DWG. NO. <i>-D-</i>	
<i>RX/F UNIT #2</i>				<i>RX/F-165-A</i>	





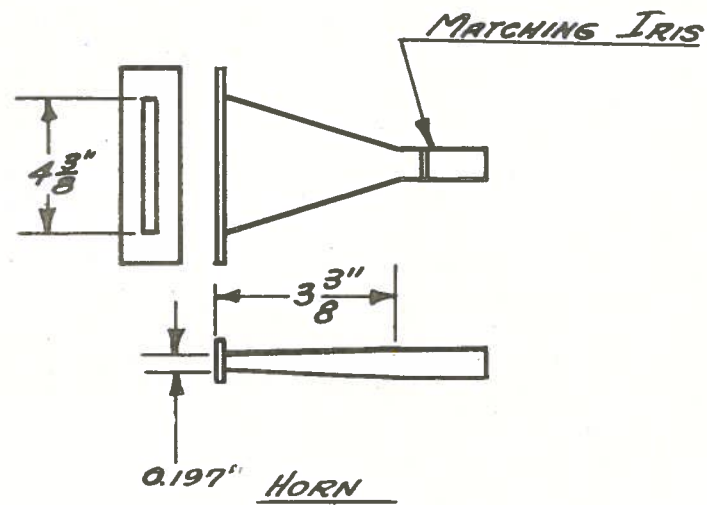
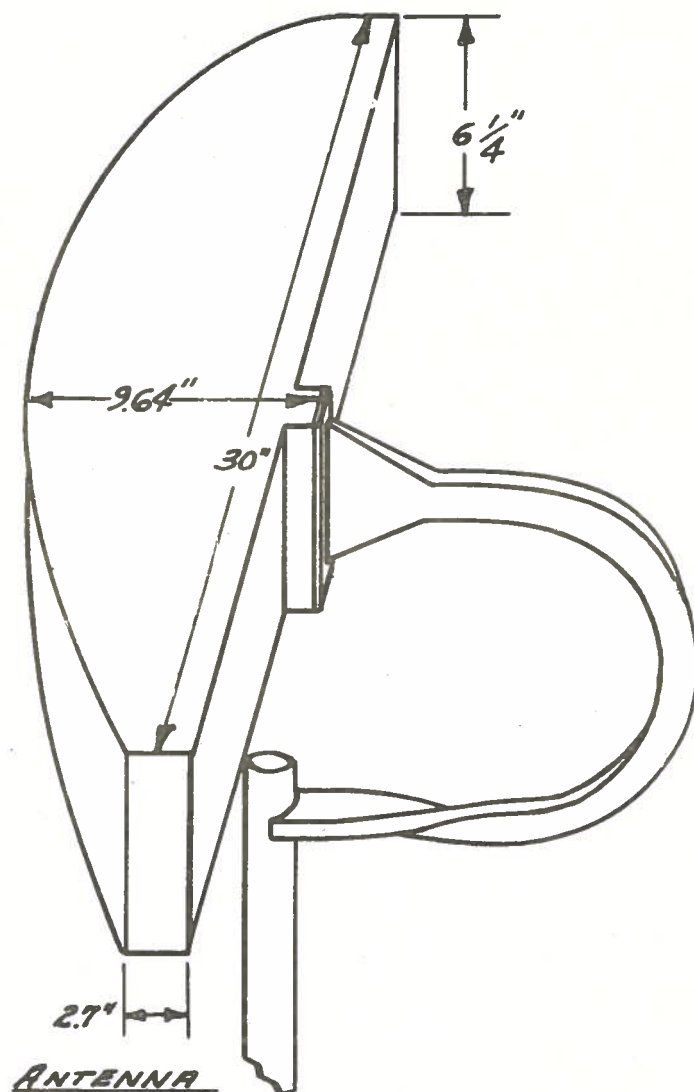


[illegible]

ONE HEATER GROUNDED IN  
THIS CHASSIS.

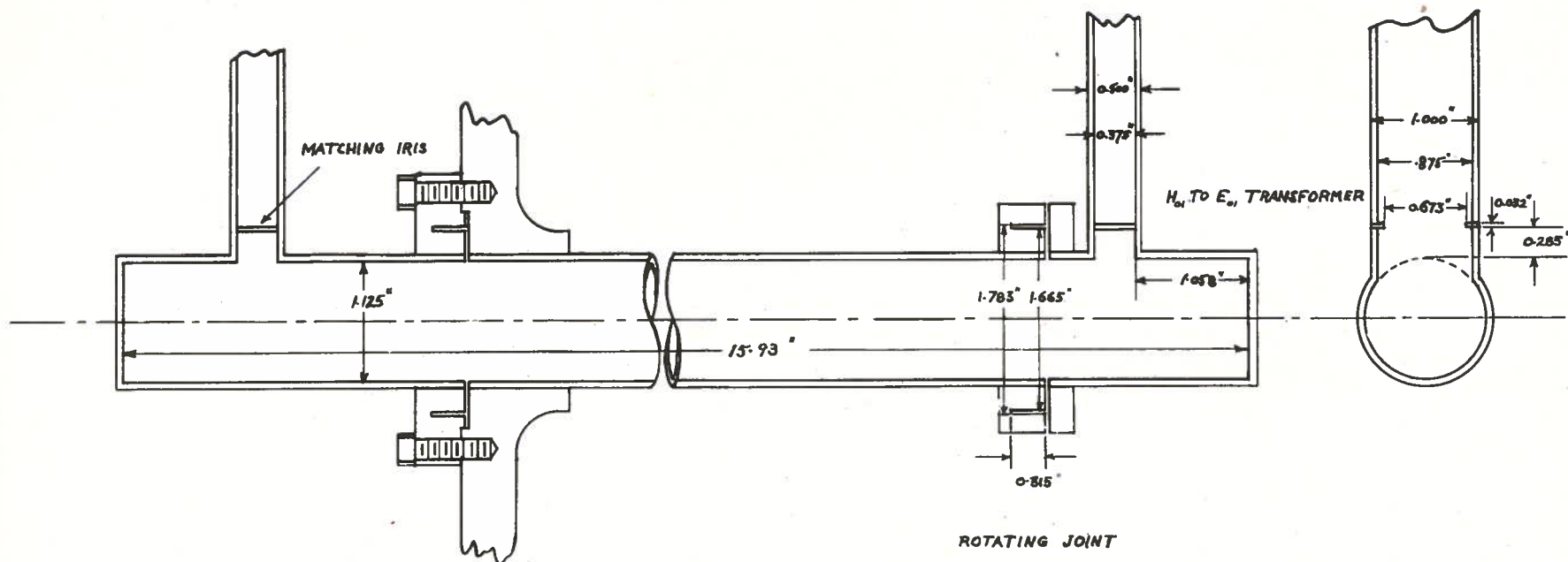
ITEM	PART NO.	Q'ANT.	MAT'L	DESCRIPTION
DRAWN BY	D. L. S.		DATE	14-4-43
CHECKED	H. W. M.		DATE	SCALE
ENG. APPROV.	A. M. M.		DATE	16-4-43
				FINISH
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA				
NAME	ESSENTIAL CIRCUITS OF P.P.I.			DWG. No.
	UNIT FOR RXIF TYPE #2			D-CA RX/F-167-B





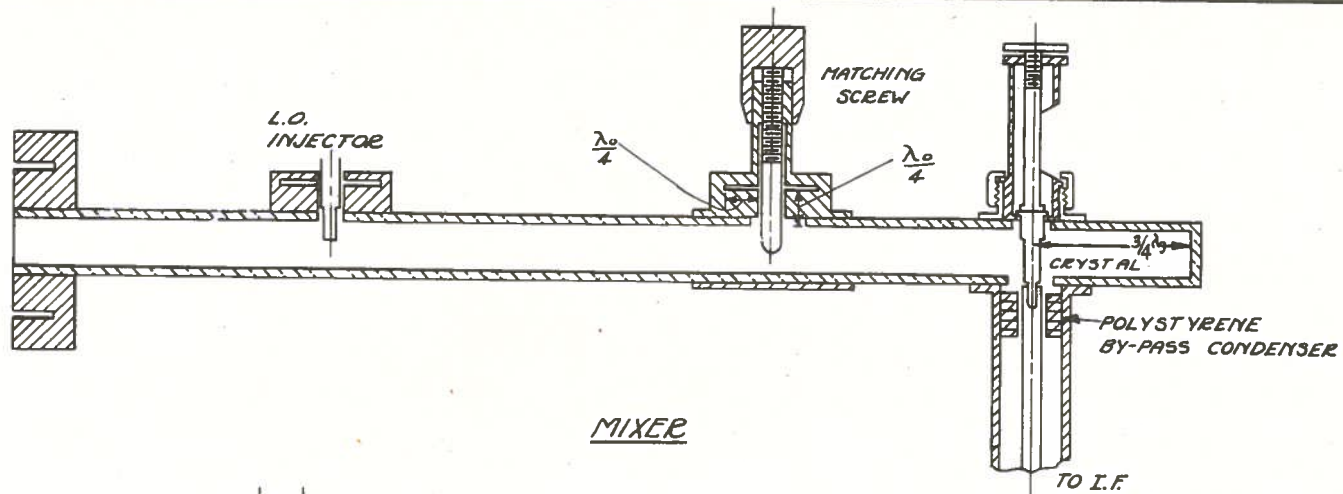
ITEM	PART NO.	QUAN.	MAT'L	DESCRIPTION	
DRAWN BY <i>E.H.B.</i>		DATE <i>FEB 21, '44</i>		SUPERSEDES	
CHECKED		DATE		SCALE	
ENG. APPROV. <i>PR</i>		DATE		FINISH.	
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA					
NAME <i>SKETCH OF RX/F ANTENNA ASSEMBLY. RX/F - 168A</i>				DWG. NO. <i>1A-A</i>	



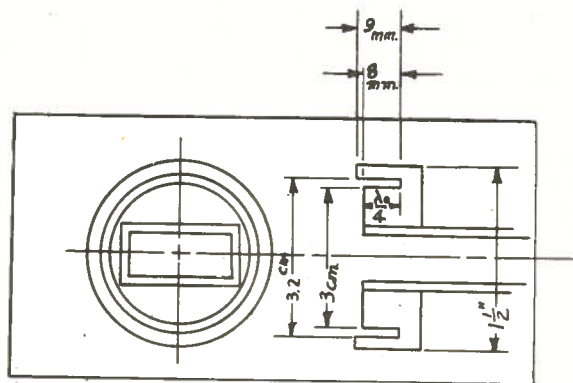


ITEM	PART NO.	QUAN.	MATL.	DESCRIPTION
DRAWN BY	C.H.M.	DATE	18/1/44	SUPERSEDES
CHECKED	H.W.M.	DATE		SCALE
ENG. APPROV.	R.M.C.	DATE		FINISH.
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA				
NAME RX/F ROTATING COUPLER.				DWG. NO. RX/F-169-B.

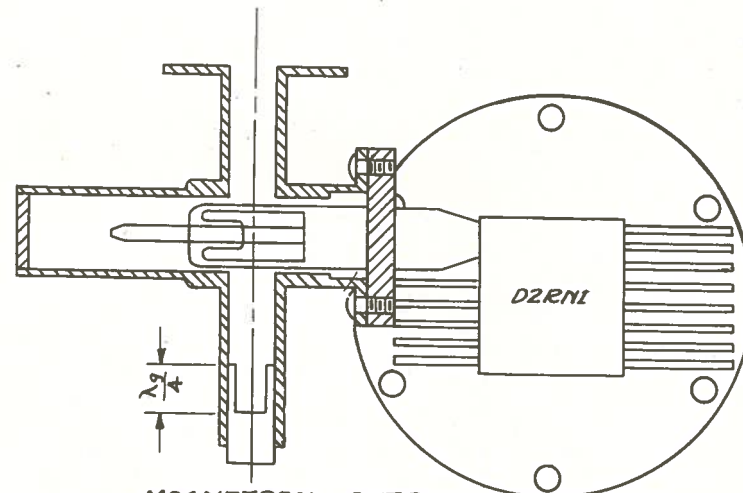




MIXER



WAVE GUIDE COUPLER

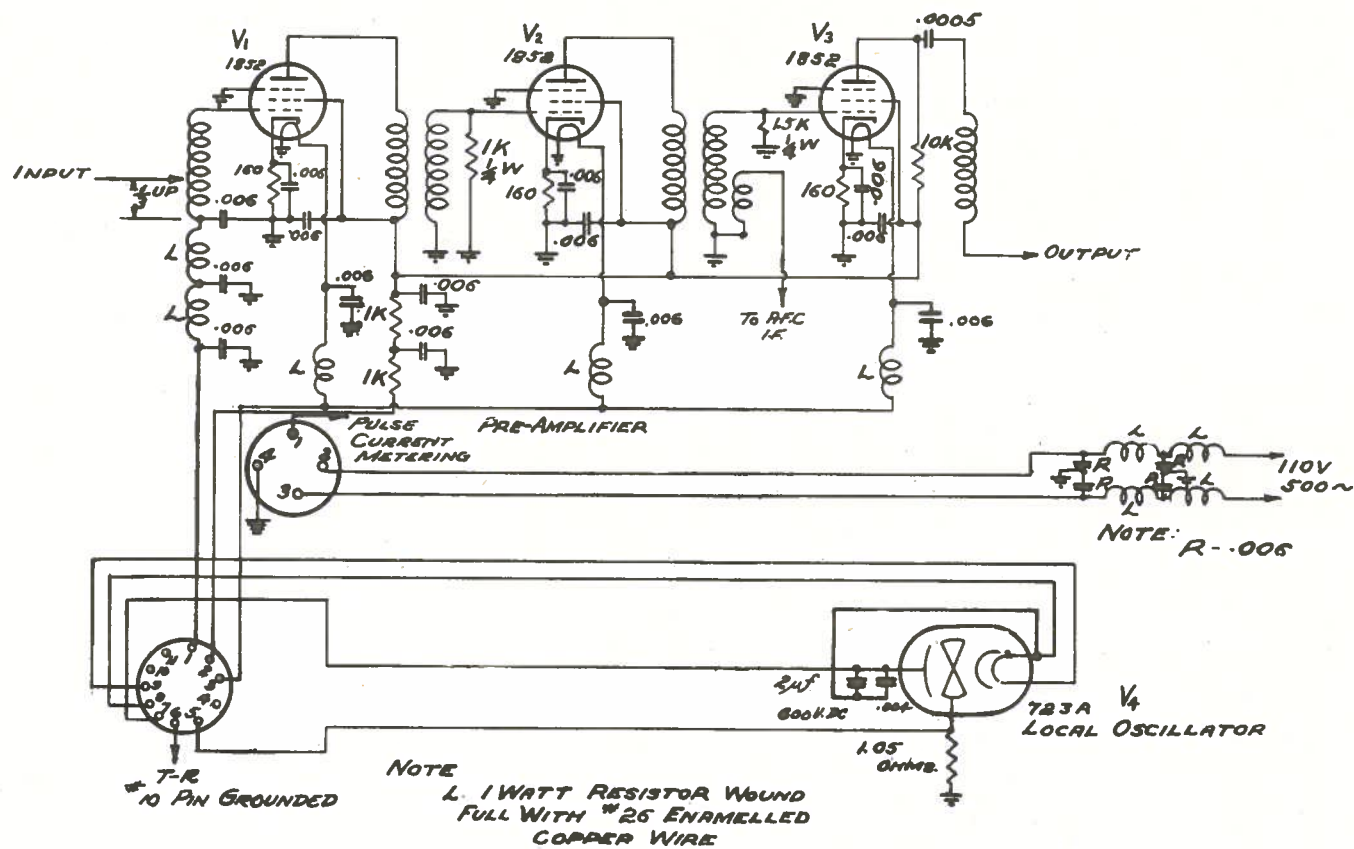


MAGNETRON OUTPUT

O.K. For P.R.B. 122 or '14

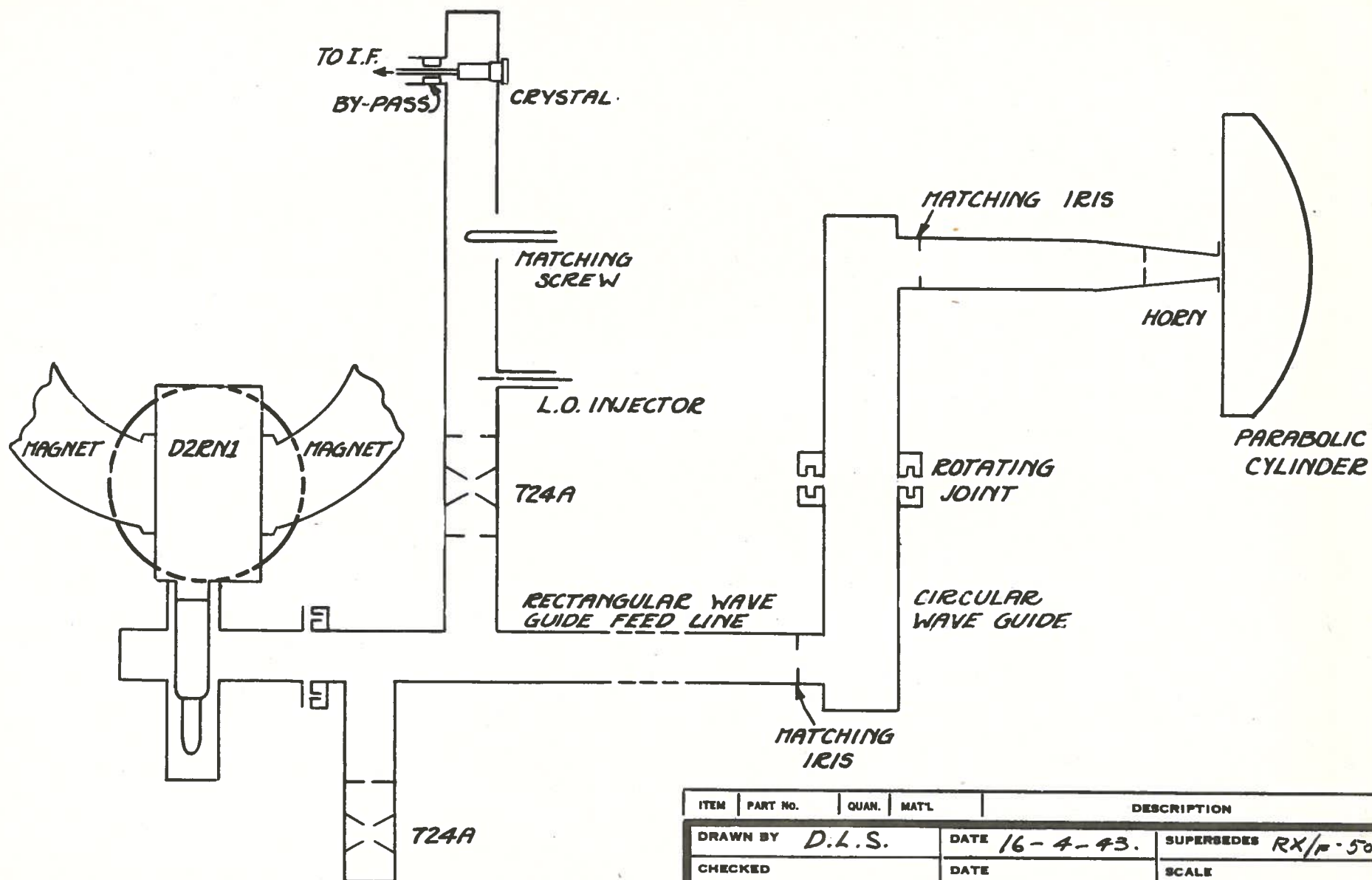
ITEM	PART NO.	QUAN.	MATL.	DESCRIPTION
DRAWN BY D.L.S.		DATE 16-4-43		SUPERSEDES
CHECKED		DATE		SCALE
ENG. APPROV. W.D.N.		DATE		FINISH.
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA				
NAME <u>MAGNETRON- OUTPUT AND MIXER.</u>				DWG. NO. <u>A-L.</u>
				RX/1-170-B





ITEM	PART NO.	QTY.	MATL.	DESCRIPTION
DRAWN BY	J.R.D.	DATE	17/4/43	SUPERSEDES
CHECKED	H.W.H.	DATE		SCALE
ENG. APPROV.	G.R.M.	DATE		FINISH
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA				
NAME	PRE-AMPLIFIER			DWG. NO. (R-5 R-IF RX/F-171-B

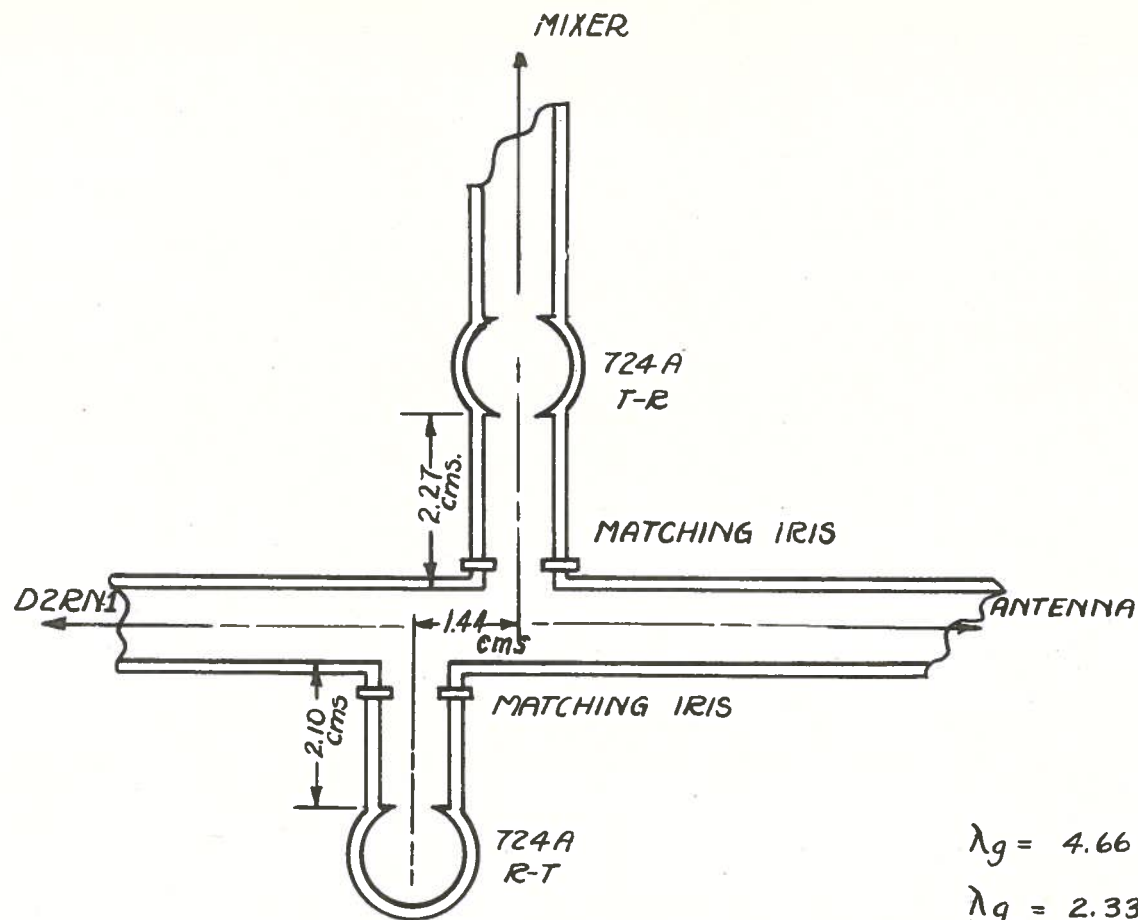




ITEM	PART NO.	QUAN.	MAT'L	DESCRIPTION
DRAWN BY <i>D.L.S.</i>		DATE <i>16-4-43.</i>		SUPERSEDES <i>RX/P-50-B</i>
CHECKED		DATE		SCALE
ENG. APPROV. <i>W.D.N.</i>		DATE		FINISH.
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA				
NAME <i>SCHEMATIC DIAGRAM OF R.F. SYSTEM</i>				DWG. NO. <i>-A-</i> <i>RX/F-172-A</i>







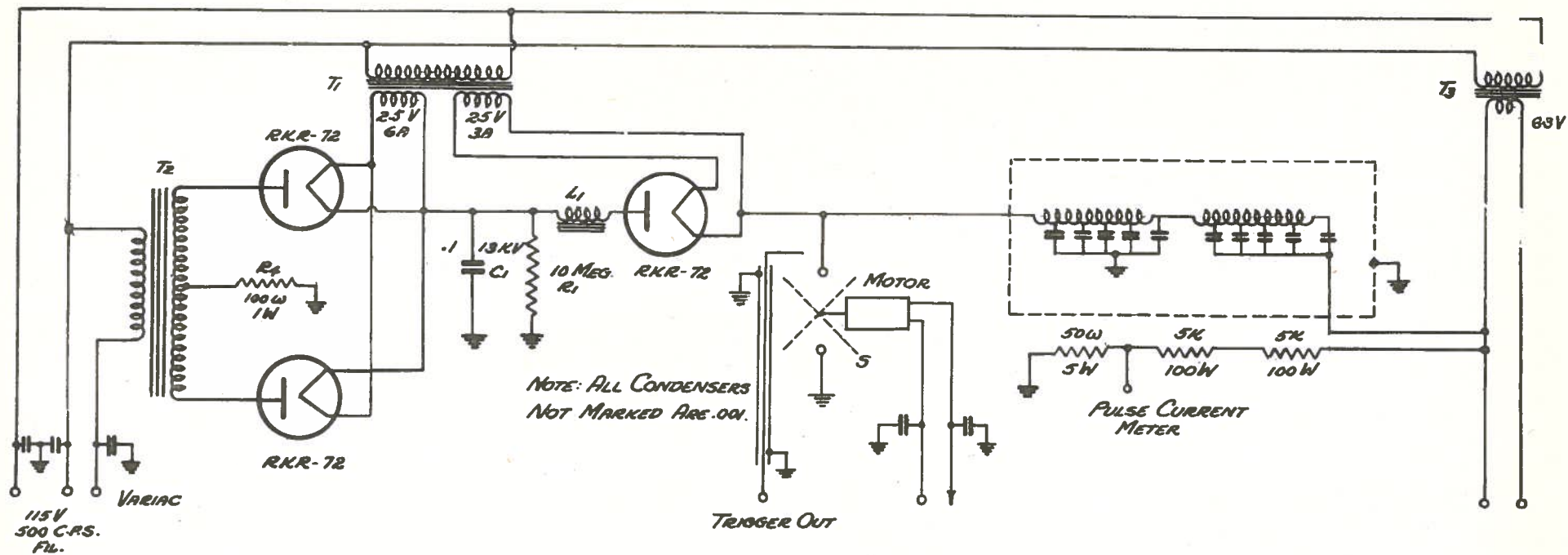
$$\lambda_g = 4.66 \text{ cm.}$$

$$\frac{\lambda_g}{2} = 2.33 \text{ cm.}$$

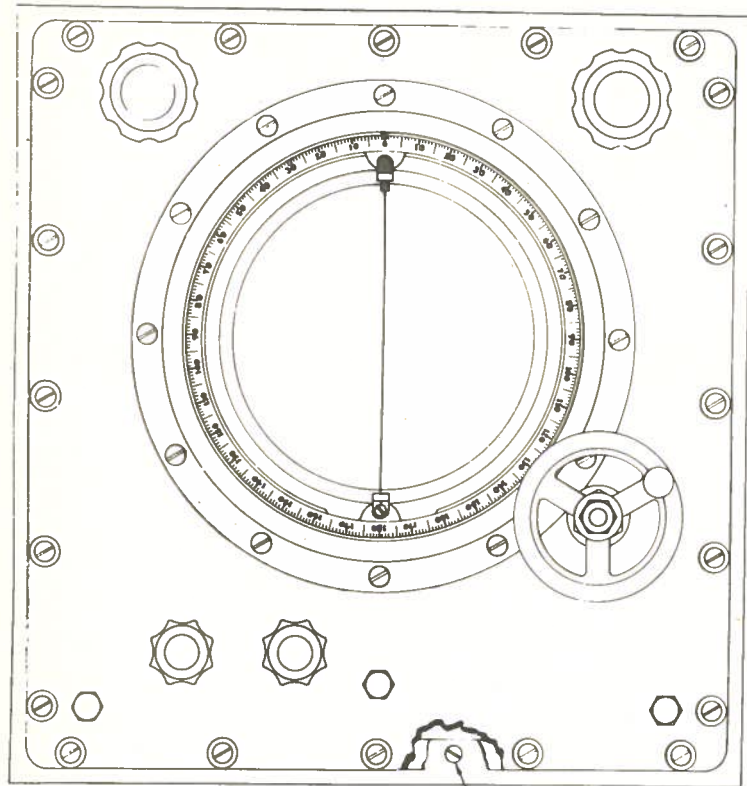
$$\frac{\lambda_g}{4} = 1.17 \text{ cm.}$$

ITEM	PART NO.	QUAN.	MAT'L	DESCRIPTION
DRAWN BY D. L. S.		DATE 15-4-43		SUPERSEDES
CHECKED		DATE		SCALE
ENG. APPROV. <i>C/M</i>		DATE		FINISH.
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA				
NAME				DWG. NO.
DUPLIXING SYSTEM				R-RF
				RX/F-174-A.

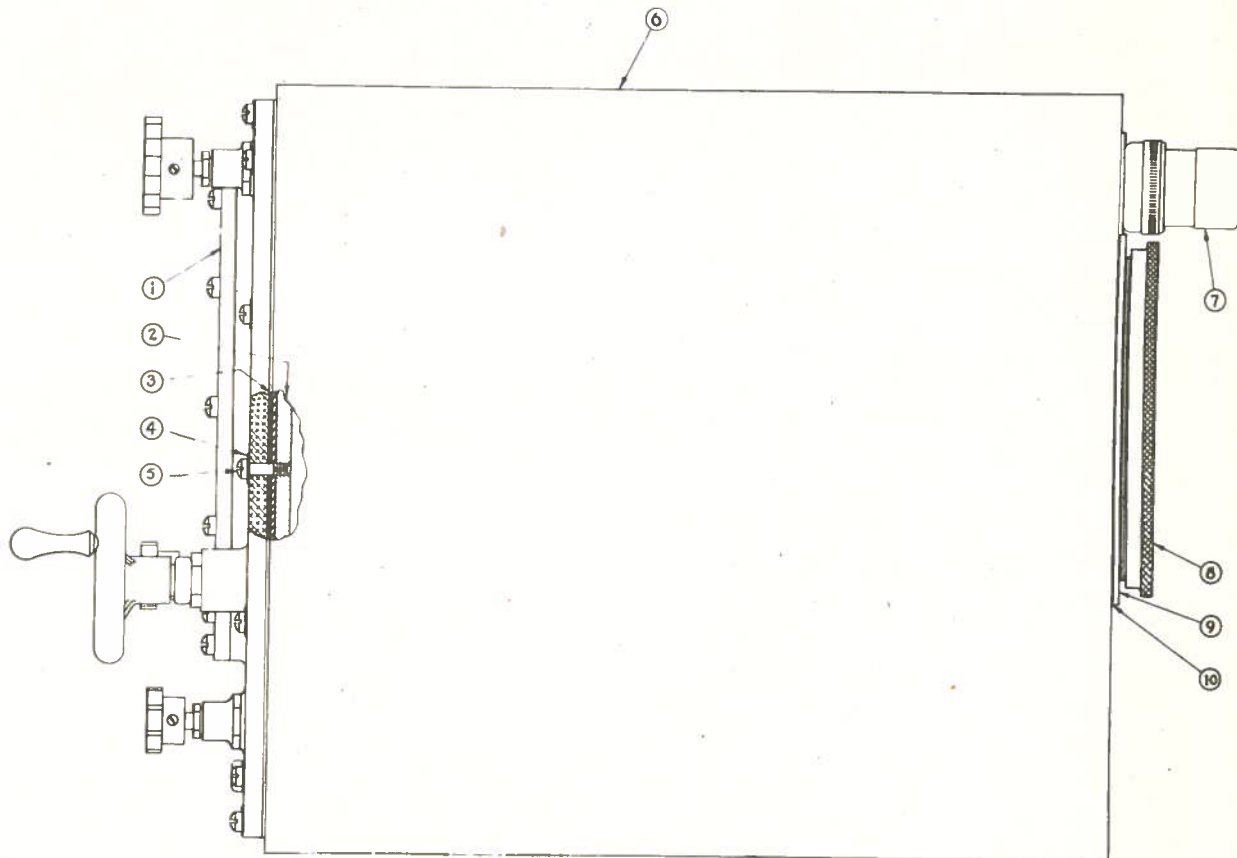




ITEM	Part No.	QUAN.	MAT'L	DESCRIPTION
DRAWN BY	DC	DATE	4-2-44	SUPERSEDED
CHECKED	H.W.M.	DATE		SCALE
ENG. APPROV.	W.R.W.	DATE		FINISH
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA				
NAME	RX/F PULSER AND CONTROL UNIT			DWG. NO. 72
	RX/F-176-B			



11

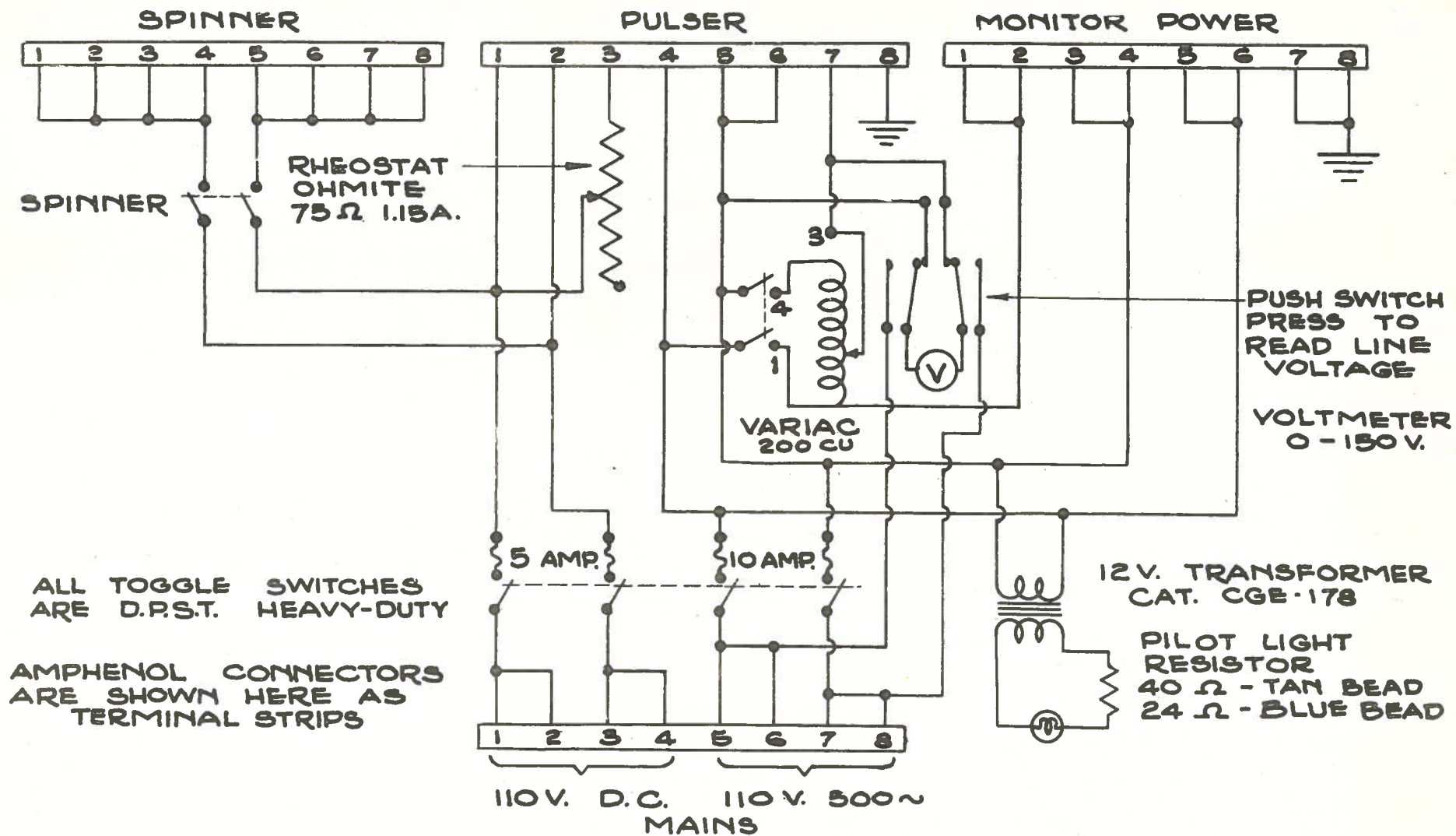


NO.	QTY	QTY	DESCRIPTION
1	124-B	1	WATERPROOF PFI ASSEMBLY
2	104-B	1	FRONT PANEL MOUNTING RING
3	221-B	1	GASKET FOR FRONT PANEL MOUNTING RING
4		20	FLAT WASHER 1/2" D
5		20	10-32 FILLISTER HEAD M.S. 3/4" LONG
6	107-B	1	WATERPROOF VP1 CASE
7	51-B	1	CABLE COUPLING ASSEMBLY
8	104-A	1	HAND HOLE COVER
9	103-B	1	HAND HOLE FLANGE
10	220-B	1	HAND HOLE FLANGE GASKET
11		2	6-32 FLAT HEAD M.S. 5/16" LONG

REVISED-13-10-45 - Rm. 100-45  
AND 100-45

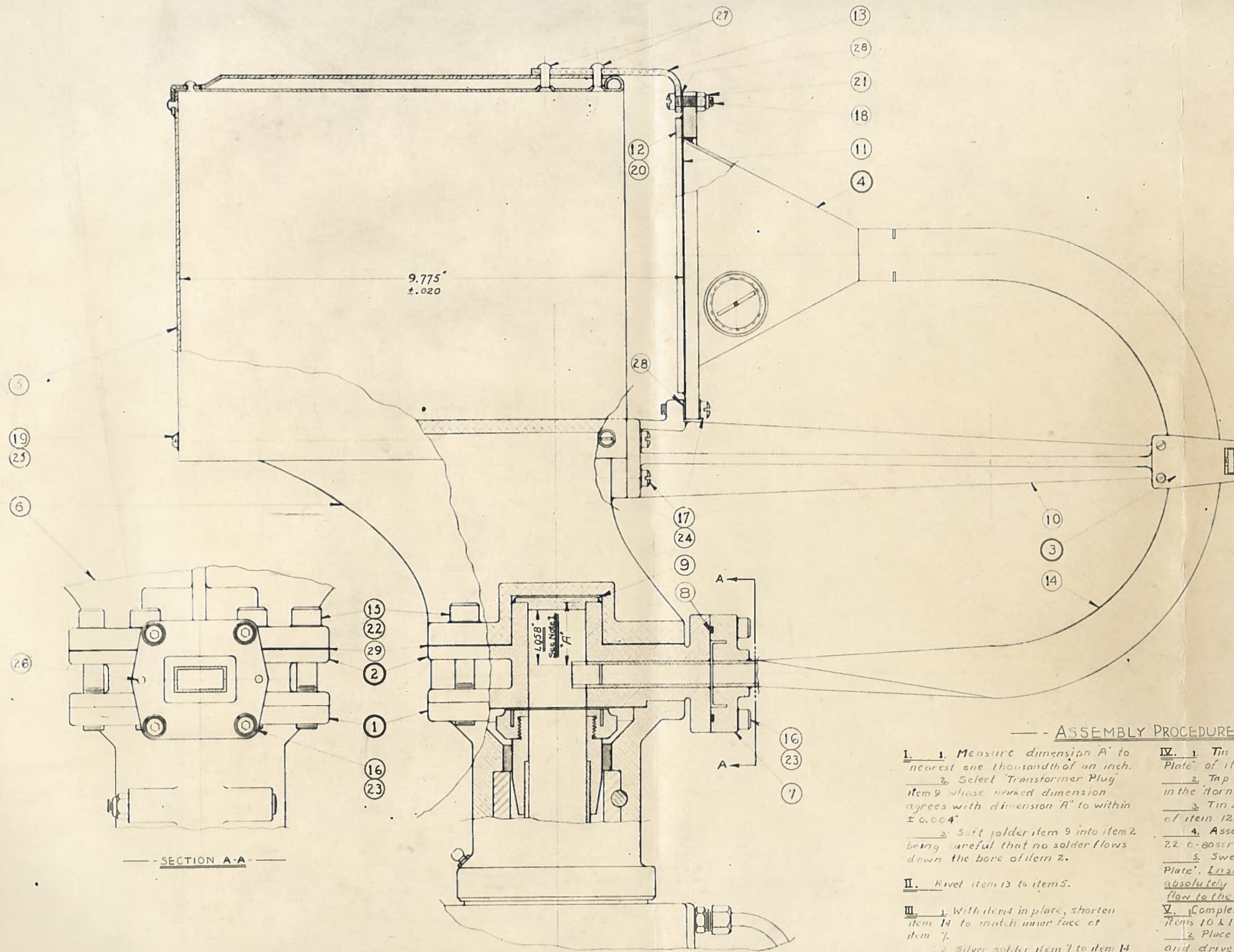
DATE	10-10-45	BY	100-45
APPROVED BY	100-45	DATE	10-10-45
REVISION	100-45	DATE	10-10-45
NATIONAL RESEARCH COUNCIL-RADIO SECTION - 100-45			
WATERPROOF PFI IN CASE			
100-45			





ITEM	PART NO.	QUAN.	MAT'L	DESCRIPTION	
DRAWN BY J.E.W.		DATE AUG. 11 - 43		SUPERSEDES	
CHECKED H.W.M.		DATE		SCALE	
ENG. APPROV. J.E.W.		DATE		FINISH.	
NATIONAL RESEARCH COUNCIL-RADIO SECTION - OTTAWA CANADA					
NAME		DWG. NO.		G-W	
POWER CONTROL PANEL		RX-F-324-A			





\* All items marked with an asterisk are to be zinc plated.

# — ASSEMBLY PROCEDURE —

- I. 1. Measure dimension A\* to nearest one thousandth of an inch.
2. Select Transformer Plug item 9 whose marked dimension agrees with dimension A\* to within  $\pm 0.004$ .
3. Soft solder item 9 into item 2 being careful that no solder flows down the bore of item 2.
- II. Rivel item 13 to item 5.
- III. 1. With item 4 in place, shorten item 14 to match inner face of item 7.
2. Silver solder item 7 to item 14 with face of item 7 at rt angles to axis of item 14.
3. Remove excess solder on recessed face of item 7.
- IV. 1. Tin & wipe smooth the Horn Face Plate\* of item 4.
2. Tap 0-80 NF-2 the 22 holes in the Horn Face Plate\*.
3. Tin & wipe smooth back face of item 12.
4. Assemble items 4, 11 & 12 with 22 0-80 screws.
5. Sweat item 12 to Horn Face Plate\*. Ensure that this joint is absolutely watertight. Solder must flow to the edge of the mica.
- V. 1. Completely assemble item 4 to items 10 & 13.
2. Place item 8 in slot of item 7 and drive in Dowel Pins\* item 26 to align flanges of items 2 & 7.
3. Tighten down the 4 cap screws so that the flanges fit tightly together.

ITEM	QTY	DESCRIPTION
29 391-A	1	Paper
28	2	St Steel 1/8" ID, 3/16" OD, 1/2" Thick Washer
27	2	Alum 1/8" ID, 3/16" OD, 1/2" Thick Washer
26	2	Brass Dowel Pin 1/16" dia, 1/2" long
25	39	Brass 0.4375" dia, 1/2" long Lock Washers
24	5	St Steel 0.525" dia, 1/2" long Lock Washers
23	4	Brass 0.4375" dia, 1/2" long Lock Washers
22	6	Steel 0.4375" dia, 1/2" long Lock Washers
21	1	Brass 10-32 Plastic Stop Nut
20	22	Brass 0-80 1/4" Lg Filistered Neck Screws
19	39	Brass 10-32 1/2" Lg Round Hd Neck Screws
18	1	St Steel 1/8" dia, 1/2" Lg Filister Hd Neck Screw
17	5	St Steel 10-32 1/4" Lg Filistered Neck Screws
16	4	Steel 1/8" dia, 1/2" Lg Round Hd Neck Screws
15	6	Steel 1/8" dia, 1/2" Lg Round Hd Neck Screws
14	362-C	1 Brass Horn - Wave guide
13	370-A	1 Alum. Horn - Top Support
12	369-A	1 Brass Horn - Window frame
11	368-A	1 Mica Horn - Window
10	371-B	1 Alum. Horn & Waveguide Support
9	374-A	1 Brass Transformer - Plug
8	373-A	1 Inconel Waveguide Coupling - Seal
7	372-A	1 Brass Waveguide Coupling
6	42-D	1 Alum. Parabolic Reflector - Casting
5	41-C	1 Alum. Parabolic Reflector - Top
4	361-B	1 Horn - Assy.
3	390-B	1 Waveguide Clamp Assy.
2	357-B	1 Upper Transformer Assy.
1	49-E	1 Antenna Gear Box Assy.

REVISION	DATE	DESCRIPTION
1	10-10-43	FEWW
2	11-11-43	FEWW
3	11-11-43	FEWW
4	11-11-43	FEWW
5	11-11-43	FEWW
6	11-11-43	FEWW
7	11-11-43	FEWW
8	11-11-43	FEWW
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26	11-11-43	FEWW
27	11-11-43	FEWW
28	11-11-43	FEWW
29	11-11-43	FEWW
30	11-11-43	FEWW

NATIONAL RESEARCH COUNCIL-RADIO SECTION - OFFICE  
RF & ANTENNA ASSY. REF-375-D